



Co-funded by the  
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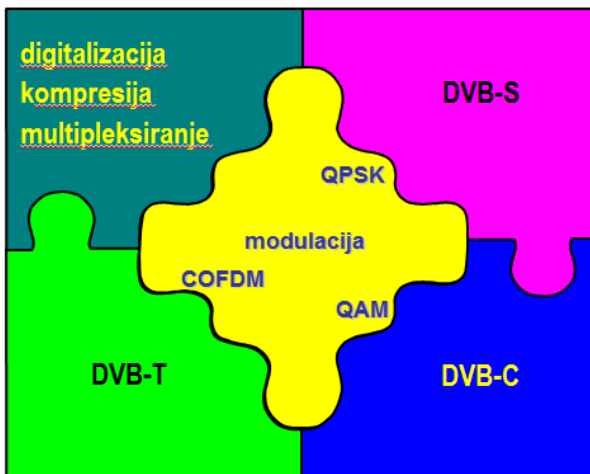
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**DBBT**

**Digital Broadcasting &  
Broadband Technologies**

# Digitalni radiodifuzni sistemi i tehnologije

## DVB



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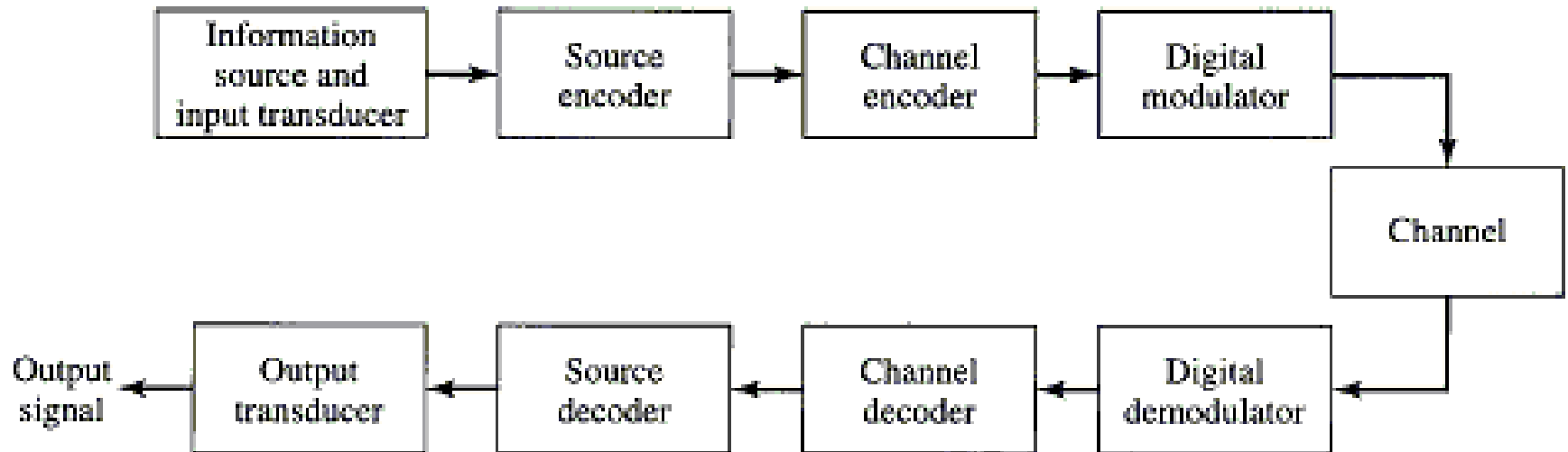
*Banja Luka, 2017.*

# Overview

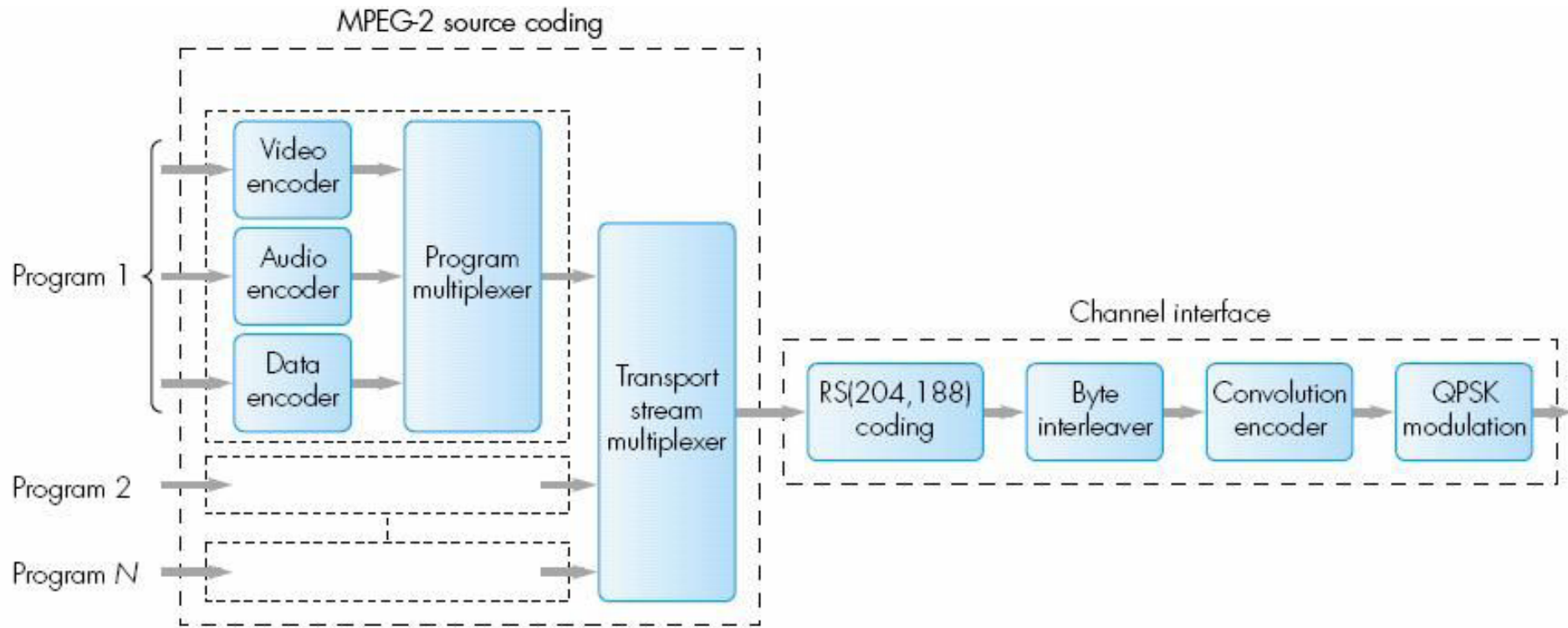
- **Digital Communication System**
- **Digital Transmission block scheme**
- **Error Correction**
- **Scrambling**
- **Digital Modulation**
- **Satellite television principles**
  - Satellites, Antennas, Encoding
- **Digital terrestrial television principles**
  - Multipath problems, Encoding, System

# Digital Communication

- Information source – MPEG stream
- Source coding – RLC, Huffman Coding
- Channel coding – Scrambling, Error Correction...
- Modulation – bits into waveform



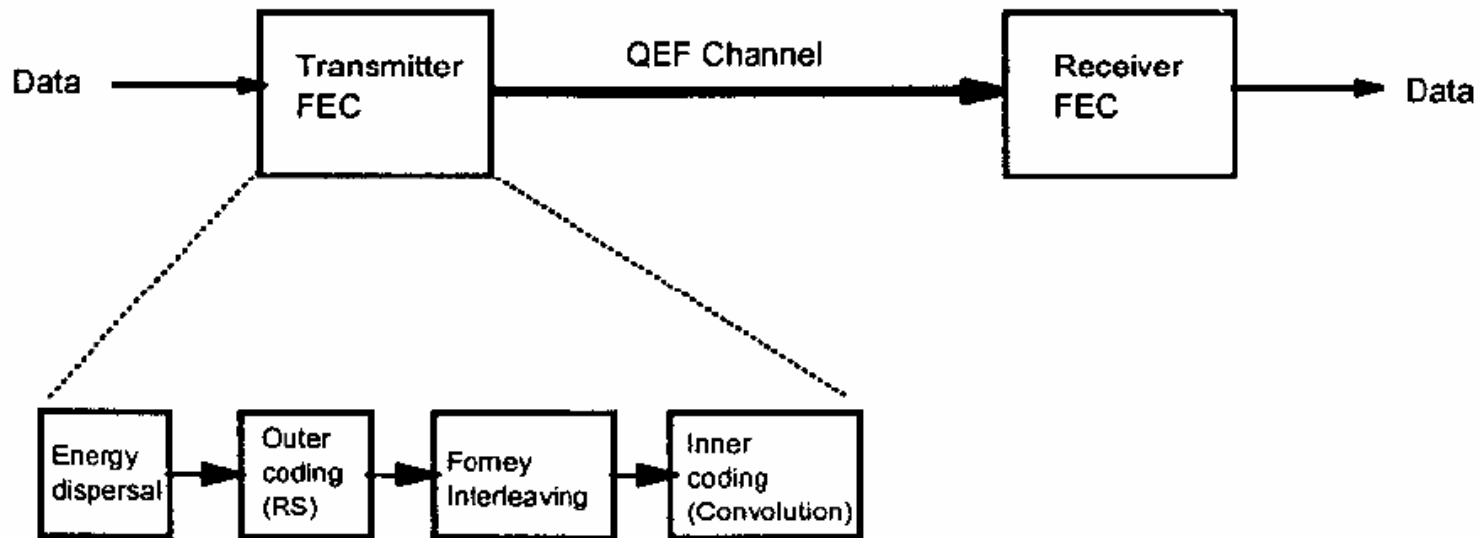
# DVB-S Transmission Scheme



Satellite digital television channel interface

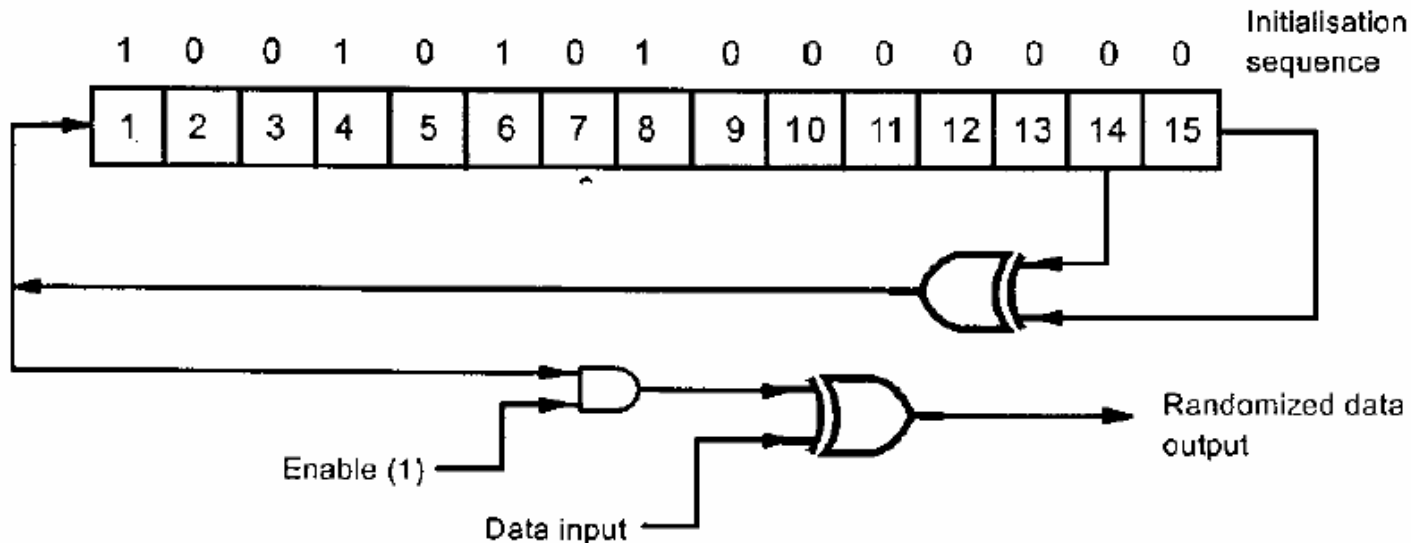
# Error Correction

- The Transport stream should be transmitted over a Quasi Error Free (QEF) channel – Bit Error Ratio :  $BER < 10^{-10}$
- TCP is resending lost packets
- Resending 3 times and voting on the correct – inefficient



# Scrambling – Energy dispersal

- Long runs of 0's or 1's generate a temporal DC coefficient
- Elementary way of encryption
- Start code and packet sync is not scrambled
- Pseudo Random Binary Sequence (PRBS)
  - Scatters bits to make energy distribution more even
  - Generated by formula :  $1+X^{14}+X^{15}$  – using shift register



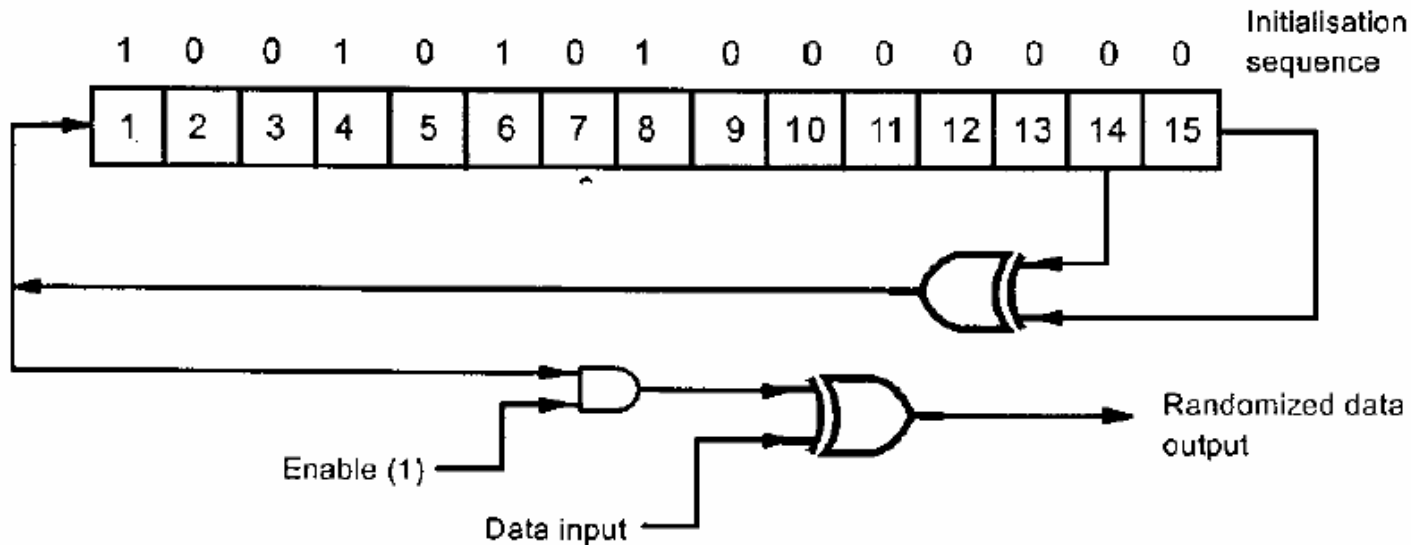
# Scrambling Example

Test:

1 1

PRBS:

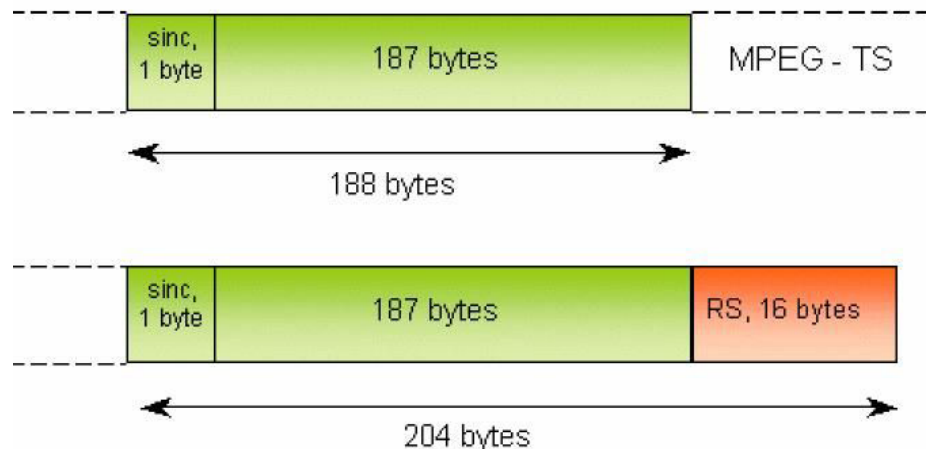
0 0 0 0 0 0 1 0 1 0 1 0 0 1 1 1 1 1 1 1 0 1 0 1 0 1 1 0 0 0 0 0 0 0





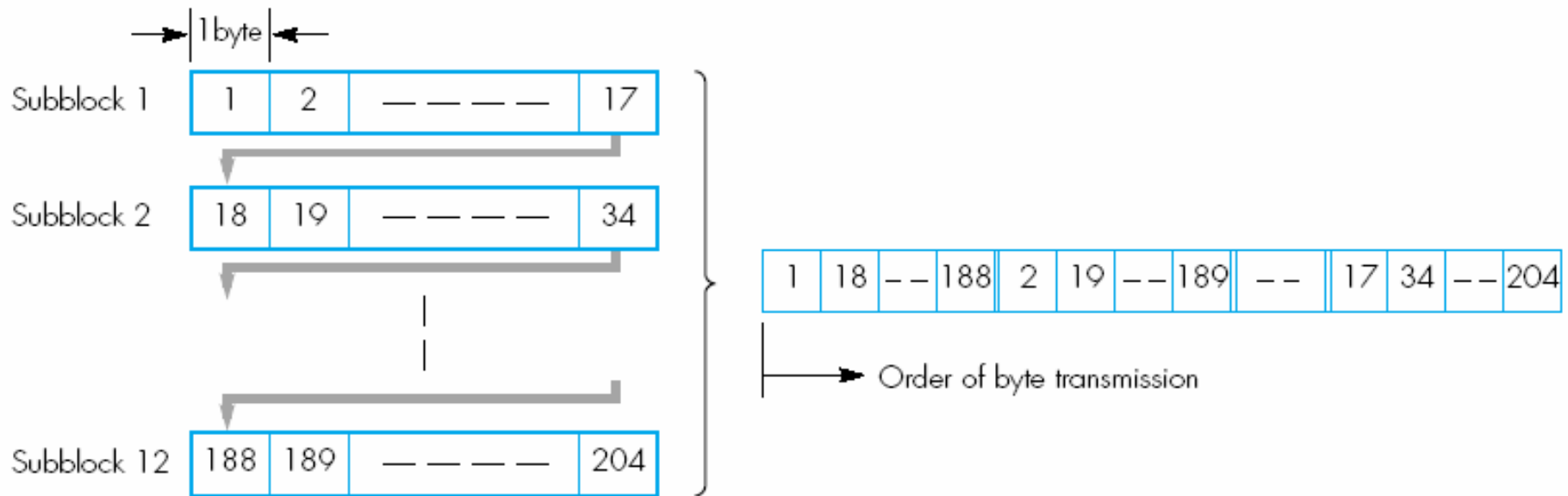
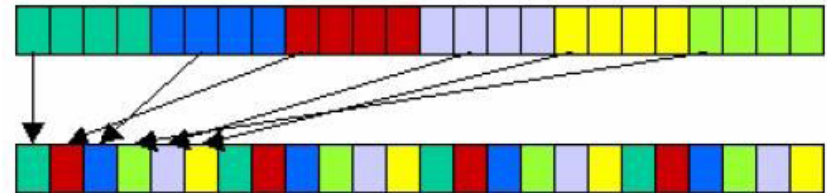
# Reed-Solomon (Outer) coding

- Block level code (works on bytes)
  - Adds extra bytes for error correction
  - Need for a block synchronization
- Characteristics:  $n$ ,  $k$ ,  $t$ 
  - $n = 204$  Final transport packet length
  - $k = 188$  Original correctable bytes
  - $t = 8$  Number of correctable bytes
- RS(204, 188) EU
- RS(207,187) USA
- Can detect  $(n-k)/2$  errors



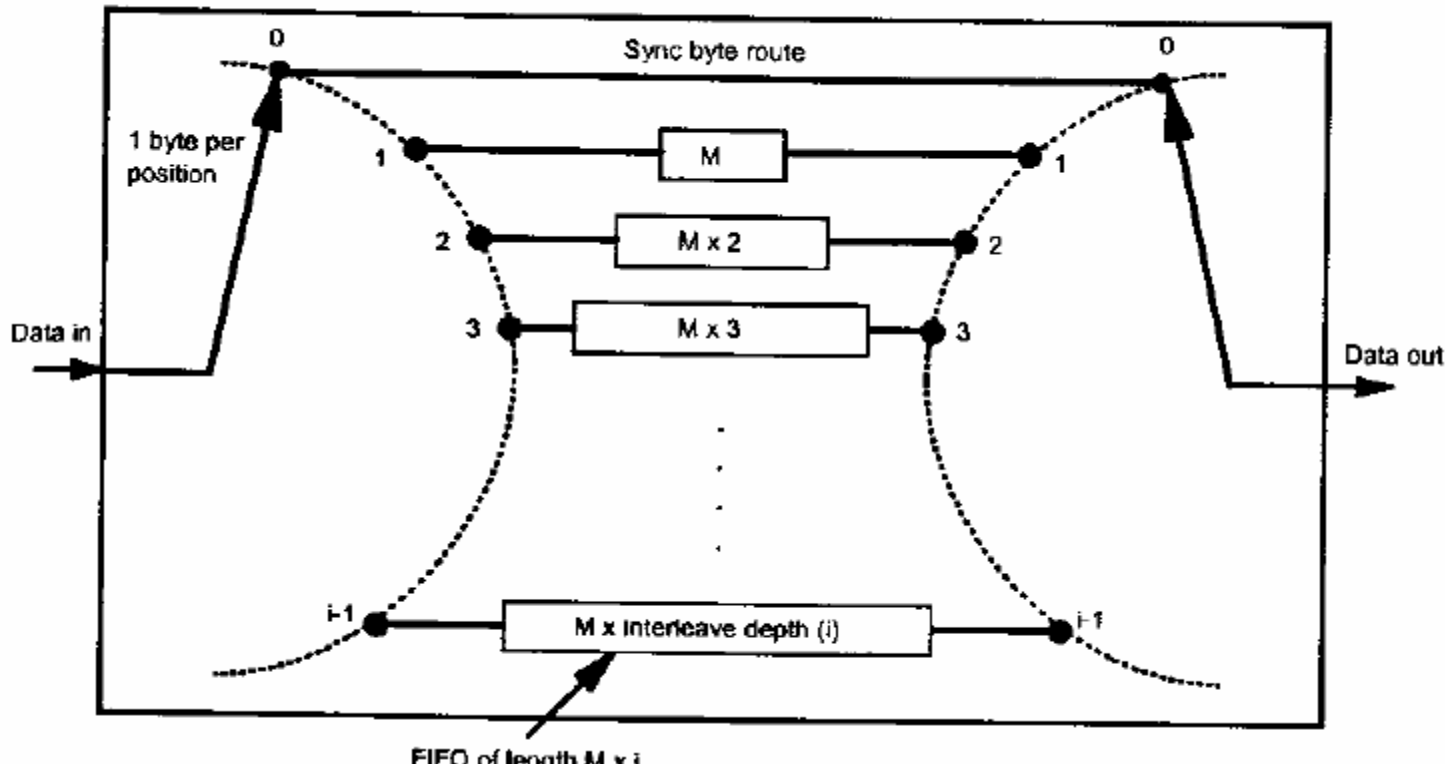
# Interleaving

- *Interleaving – breaking down very long error bursts for FEC*
- *By rearranging the order of transmission of the bytes*



# Forney (Outer) interleaving

- To spread out the error bits –more efficient outer coding
- For DVB with  $i=12$  branches and  $L=204$  packet length,  $M=(L/i) \times j$  is FIFO buffer size,  $j=0 \dots 11$



# Interleaving & Error Correction

Frequency-selective fading are indicated by **F** and errors due to time fading are indicated by **T**.

*After demodulation*

```
.rirfcFFFrgelndao  
es yr FFFAic cei c  
ernn FFFdueVet hi  
sn dr FFF it ner T  
TTTTTFFFTTTTTTTTT  
nearFFFdtsa onre  
eiefFFFriixrc ui  
l ndhqFFFrtb lodml  
vnh FFFc sama ot  
yrtoezFFFeec nimtf  
geianvFFFnateeea i  
rniitFFFah.grseat  
o-eyaFFFseuiott b  
emoiFFFrdni nr b  
n ennFFFvbl ient  
rpoixFFFudarbedee
```

*After de-interleaving*

```
This is aF ExampFe  
of text cFnFainiFg  
error burFtF TotF  
in columnF FndTiFT  
lines cauFeF by F  
TTannel sFlPctiTF  
fading. AFTPr deF  
interTeavFnF in F  
time anT FrFquenFy  
theTe errFrFThavF  
been randFmFzed Fs  
indiTaTedeF FfterF  
error corFeFtionF  
by a VitTFbF de-F  
coder theFoFiginFl  
teTt is rFcFvereF.
```

*After error correction*

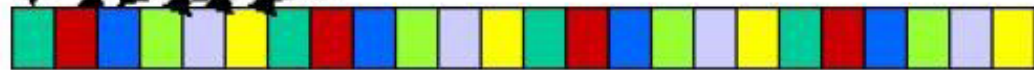
```
This is an example  
of text containing  
error bursts both  
in columns and in  
lines caused by  
channel selective  
fading. After de-  
interleaving in  
time and frequency  
these errors have  
been randomized as  
indicated. After  
error correction  
by a Viterbi de-  
coder the original  
text is recovered.
```

# Decoding with Errors

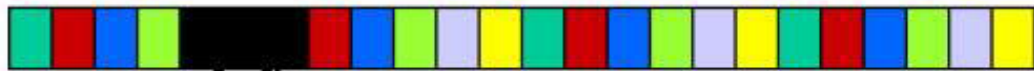
- Interleaving



- No Error



- Error burst



- Deinterleaving



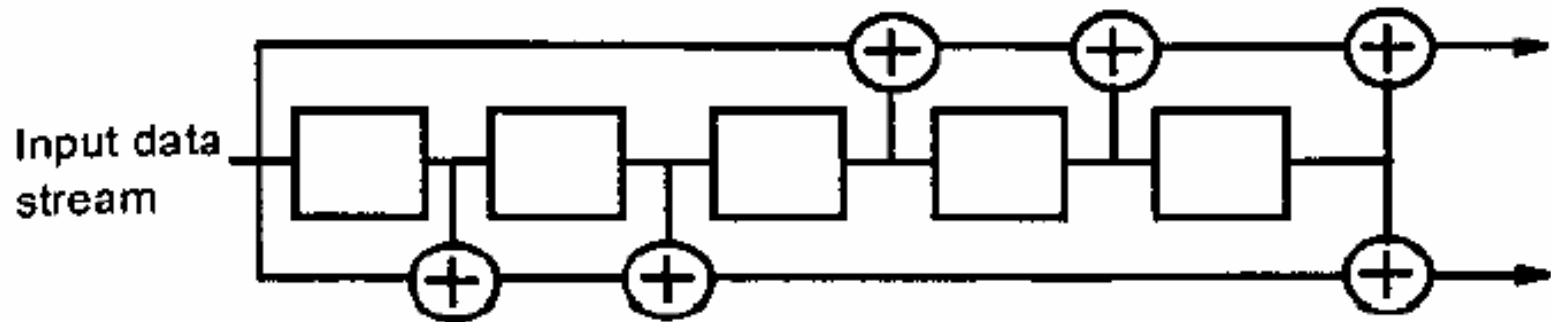
- Viterbi and Reed-Solomon



# Viterbi (Inner) Coding

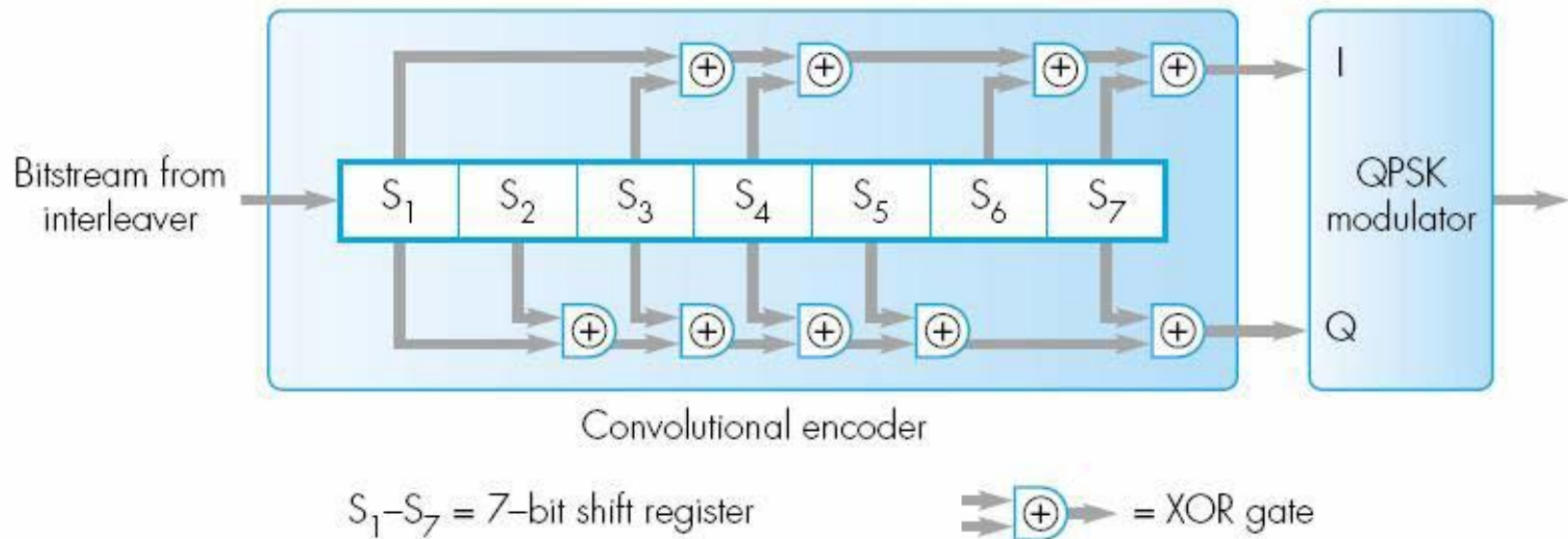
- Convolution Code – bit level
- Memory and computationally more efficient
- 2-3 data streams are produced – redundancy
- Codes system states rather than signals involved
- Calculates the probabilities of state transitions
- Trellis diagram: Decoding at the receiver side

133/171



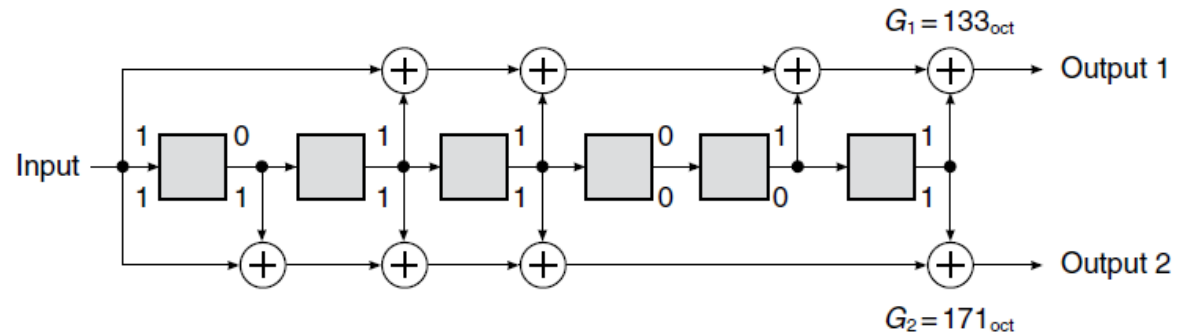
# Channel Encoding

- *Convolutional encoder*
- *Hamming distance*
- *Puncturing*



# Channel Encoding

- *Convolutional encoder*
- *Hamming distance*
- *Puncturing*



**Figure 6.8** Principle diagram of the DVB-S convolutional coder.

**Table 6.1** Basic parameters of the DVB convolutional code.

Parameter	Abbreviation	Value
Code rate	$R_c$	1/2
Constraint length	$K$	7
First polynomial generator	$G_1$	171 <sub>oct</sub>
Second polynomial generator	$G_2$	133 <sub>oct</sub>
Free distance	$d_{\text{free}}$	10



# Puncturing

- Code redundancy introduces more bit rate
- More bit rate – more errors!
- Shortening the code word – puncturing
- Periodical removal of the certain coded bits
- Regular pattern to be able to decode
- Not as prone to errors
- Trade of between the error resilience and redundancy

# Puncturing

**Table 6.2** DVB inner coding characteristics (derived from prETS 300 421).  
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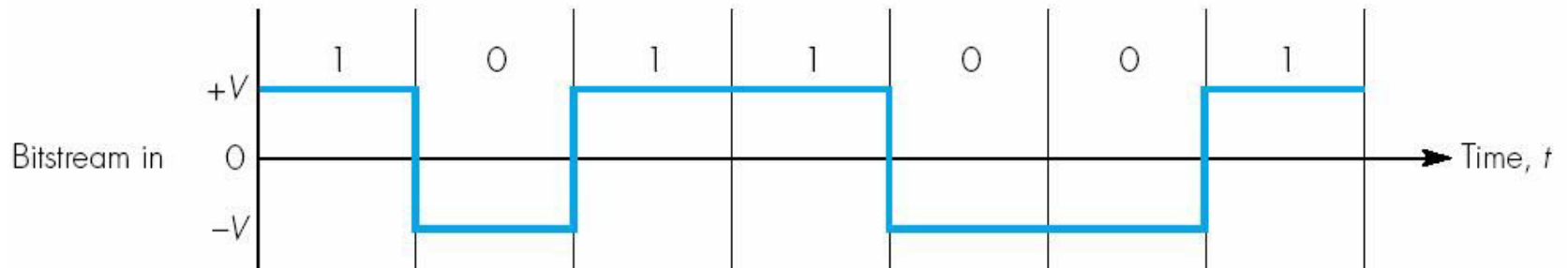
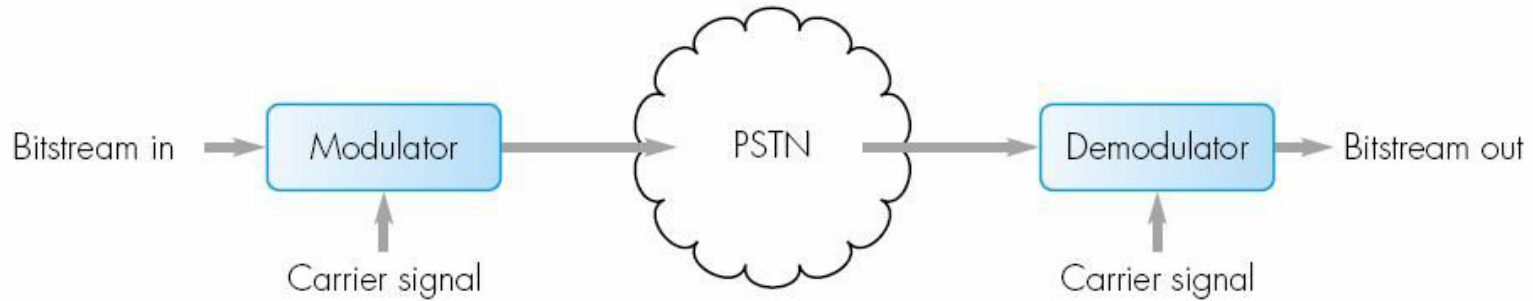
$R_c$	1/2	2/3	3/4	5/6	7/8
$d_{\text{free}}$	10	6	5	4	3
$X$	1	10 10	101	10101	1000101
$Y$	1	11 11	110	11010	1111010
$I$	$X_1$	$X_1 Y_2 Y_3$	$X_1 Y_2$	$X_1 Y_2 Y_4$	$X_1 Y_2 Y_4 Y_6$
$Q$	$Y_1$	$Y_1 X_3 Y_4$	$Y_1 X_3$	$Y_1 X_3 X_5$	$Y_1 Y_3 X_5 X_7$
$S_{\text{OFDM}}$	$X_1 Y_1$	$X_1 Y_1 Y_2 X_3 Y_3 Y_4$	$X_1 Y_1 Y_2 X_3$	$X_1 Y_1 Y_2 X_3 Y_4 X_5$	$X_1 Y_1 Y_2 Y_3 Y_4 X_5 Y_6 X_7$

*Notes:*

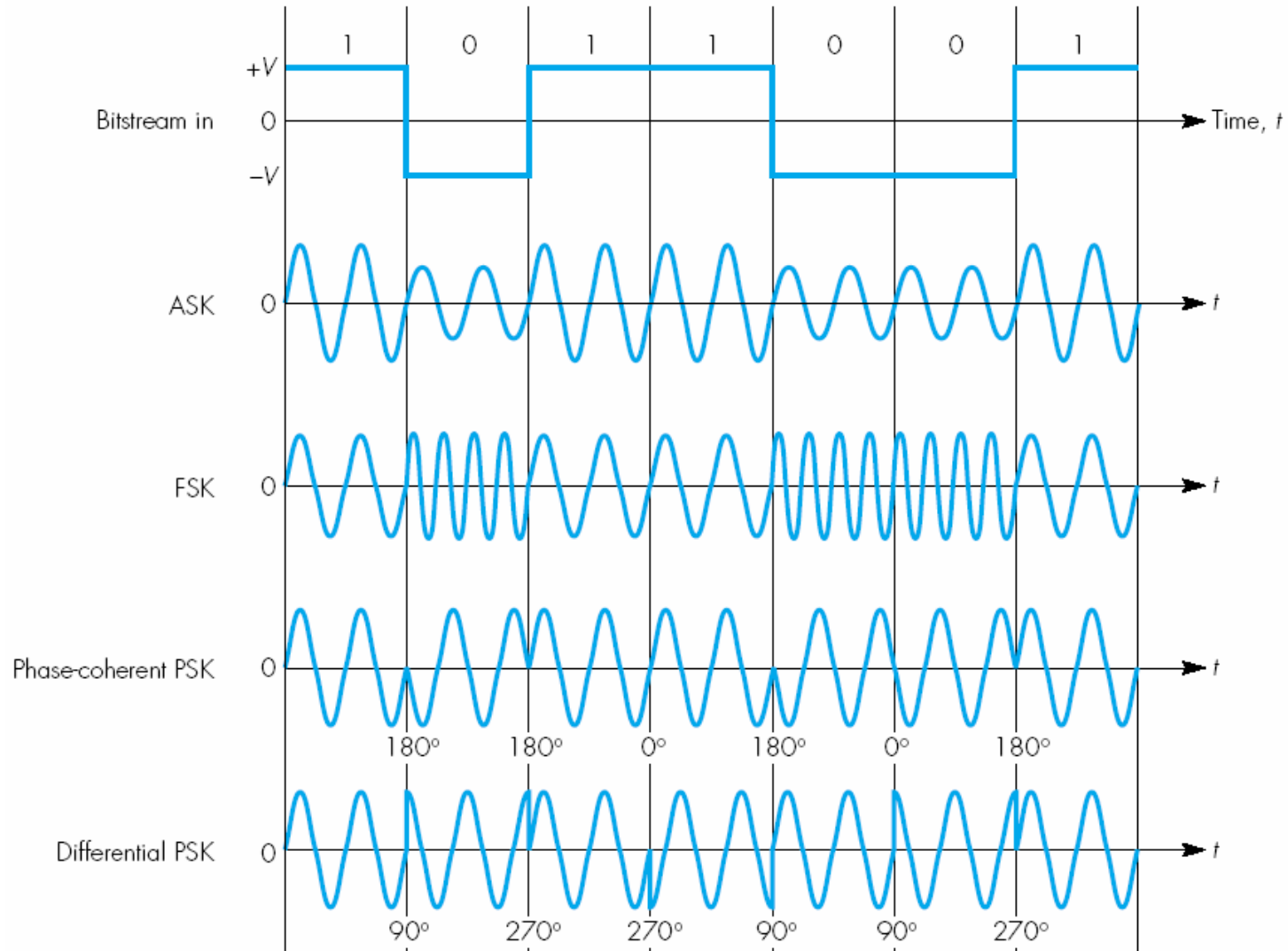
On lines  $X$  and  $Y$ , ‘0’ denotes a suppressed bit, ‘1’ denotes a transmitted bit. For terrestrial transmissions based on OFMD modulation, additional steps are required after the inner coding: serialization of the bitstream, inner interleaving and symbol mapping to adapt the bitstream format to the high number of carriers used (see Chapter 7). The last line of the table,  $S_{\text{OFDM}}$ , represents the serialized bitstream (obtained by alternating  $I$  and  $Q$  lines) applied to the inner interleaving circuit used in the case of terrestrial OFDM transmission.

# Digital Modulation

- Adapting the source signal to the transmission medium



# Digital Modulation



# Shannon's Theorem

- The capacity  $C$  of a channel

$$C = B \times \log_2(1 + \text{SNR})$$

$B$  - bandwidth

SNR - signal to noise ratio

- PAL analogue 6MHz studio: S/N = 65 dB -> 129 Mbit/s
- PAL analogue 6MHz broadcast: S/N = 21 dB -> 42 Mbit/s
- Practical for digital applications: S/N = 15 dB -> 30 Mbit/s
- Normal digital applications: 6 bps / Hz

# Modulation Notation

- Modify carriers amplitude and/or phase (and frequency)
- Constellation map: Vector notation/Complex frequency coordinates

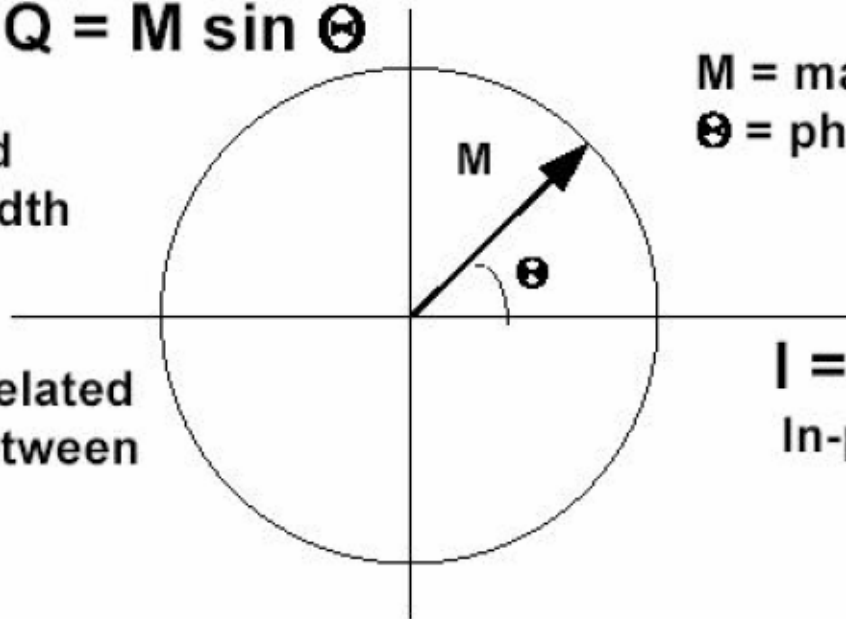
Quadrature component (carrier shifted 90°)

$$Q = M \sin \Theta$$

M = magnitude  
 $\Theta$  = phase

Densely packed  
implies bandwidth  
efficient

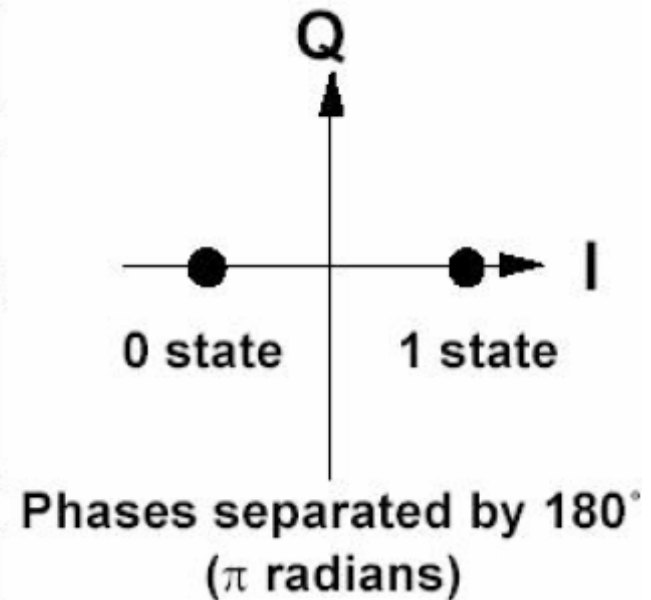
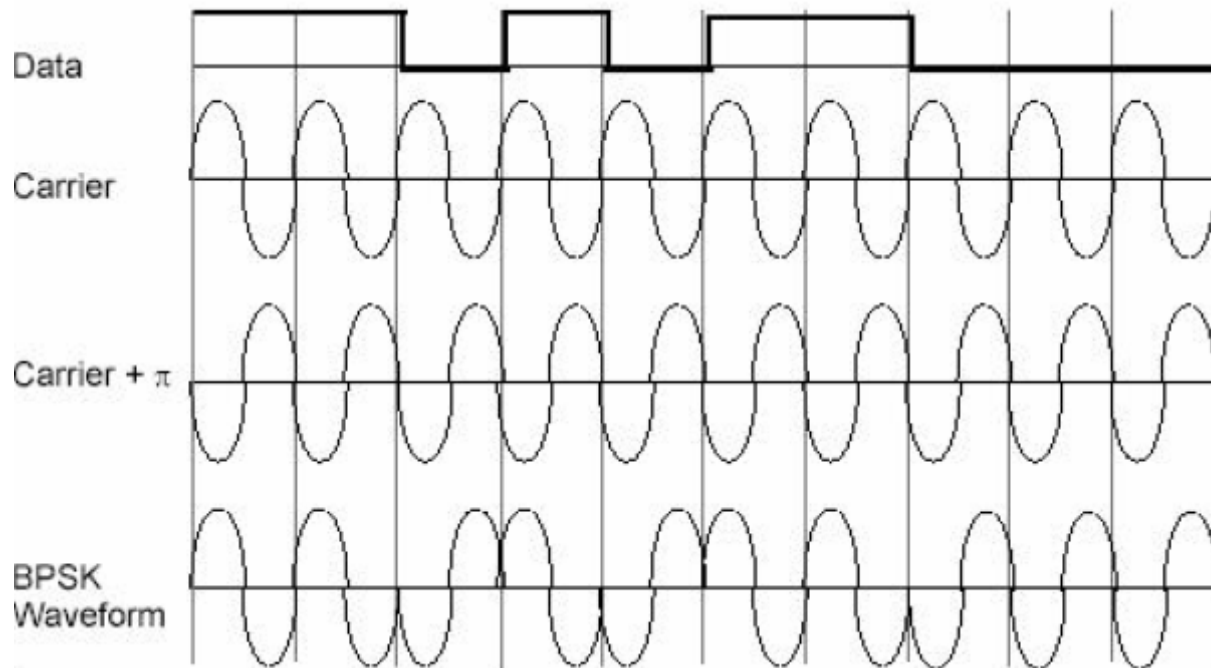
Bit error prob related  
to distances between  
closest points



$I = M \cos \Theta$   
In-phase component

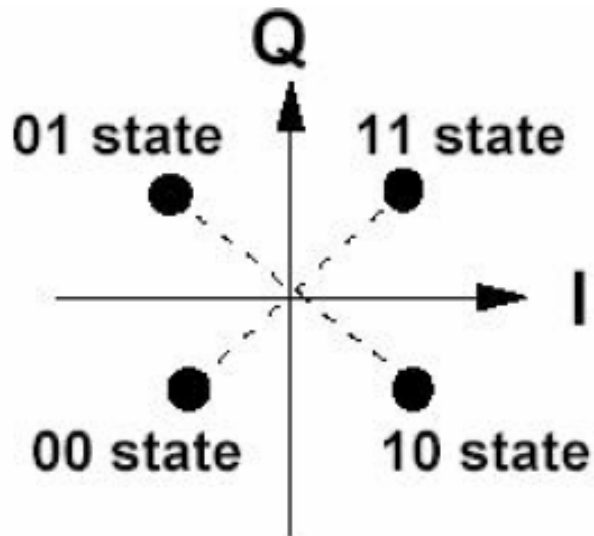
# Binary modulations

- Binary phase shift keying (BPSK)
  - Simple to implement, inefficient use of bandwidth
  - Very robust, used in satellite communications



# Phase Key Shifting (PSK)

- Quadrature Phase Shift Keying (QPSK)
  - Multilevel modulation technique: 2 bits per symbol
  - More spectral efficiency, more complex receiver



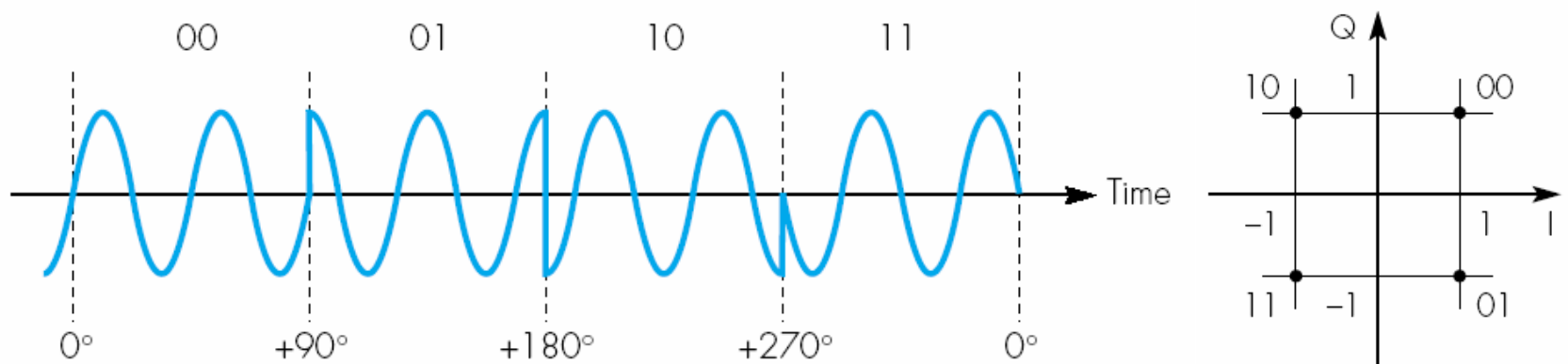
Phase of carrier:  
 $\pi/4, 3\pi/4, 5\pi/4, 7\pi/4$

Output waveform is  
sum of modulated  $\pm$   
Cosine and  $\pm$ Sine wave

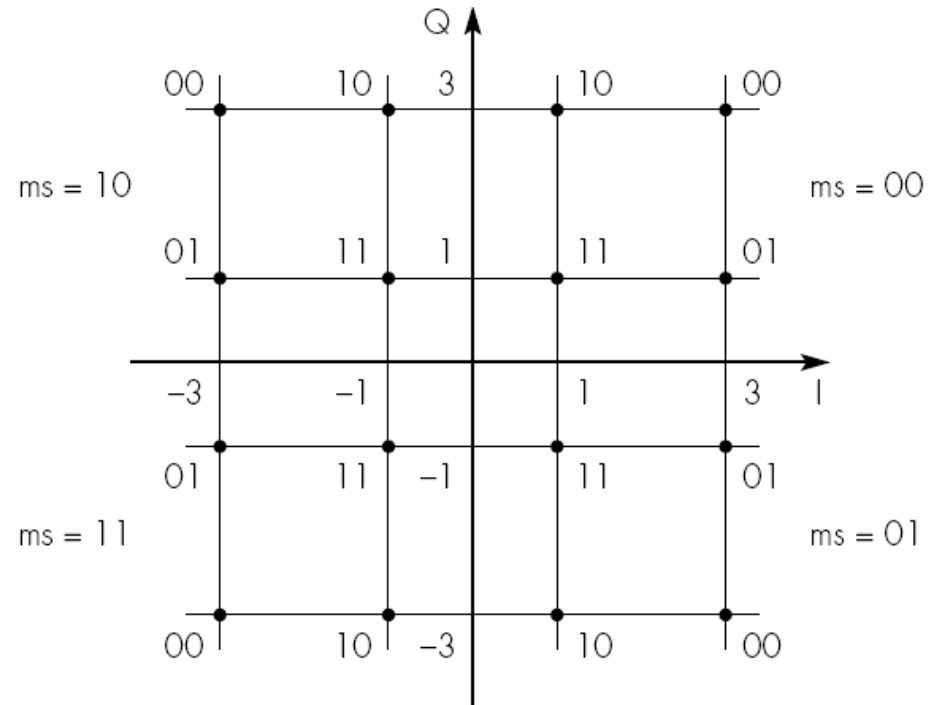
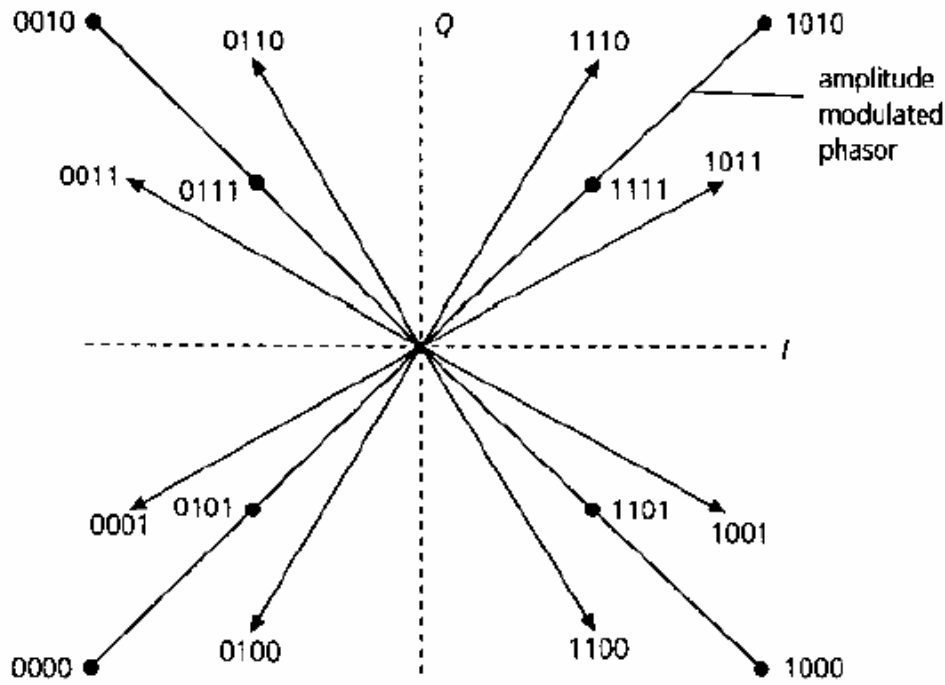


# QAM

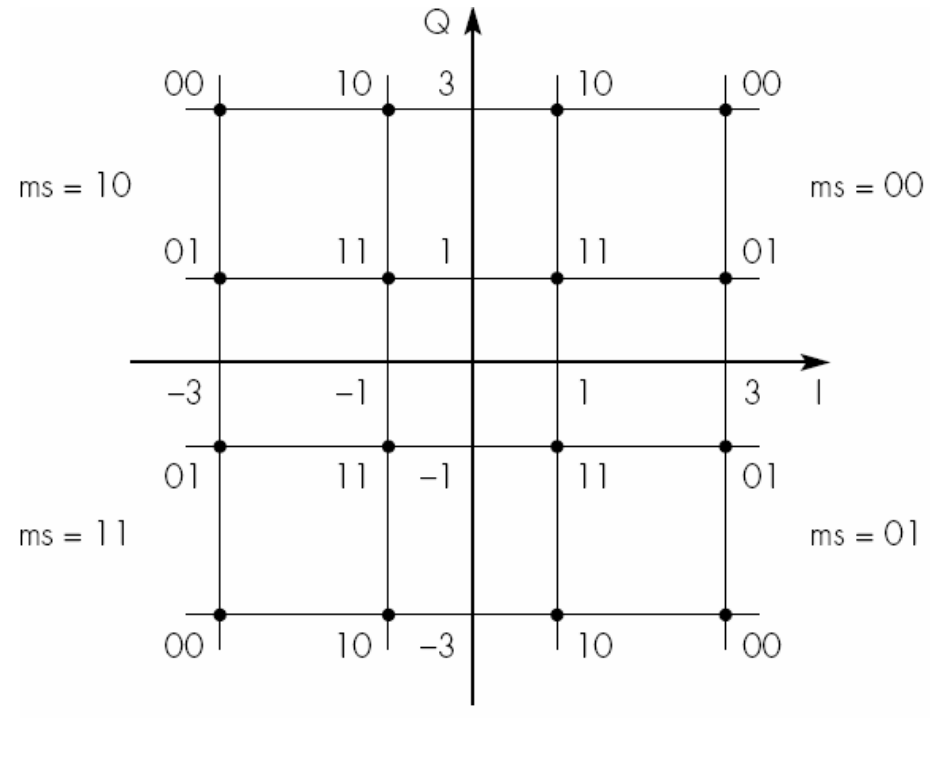
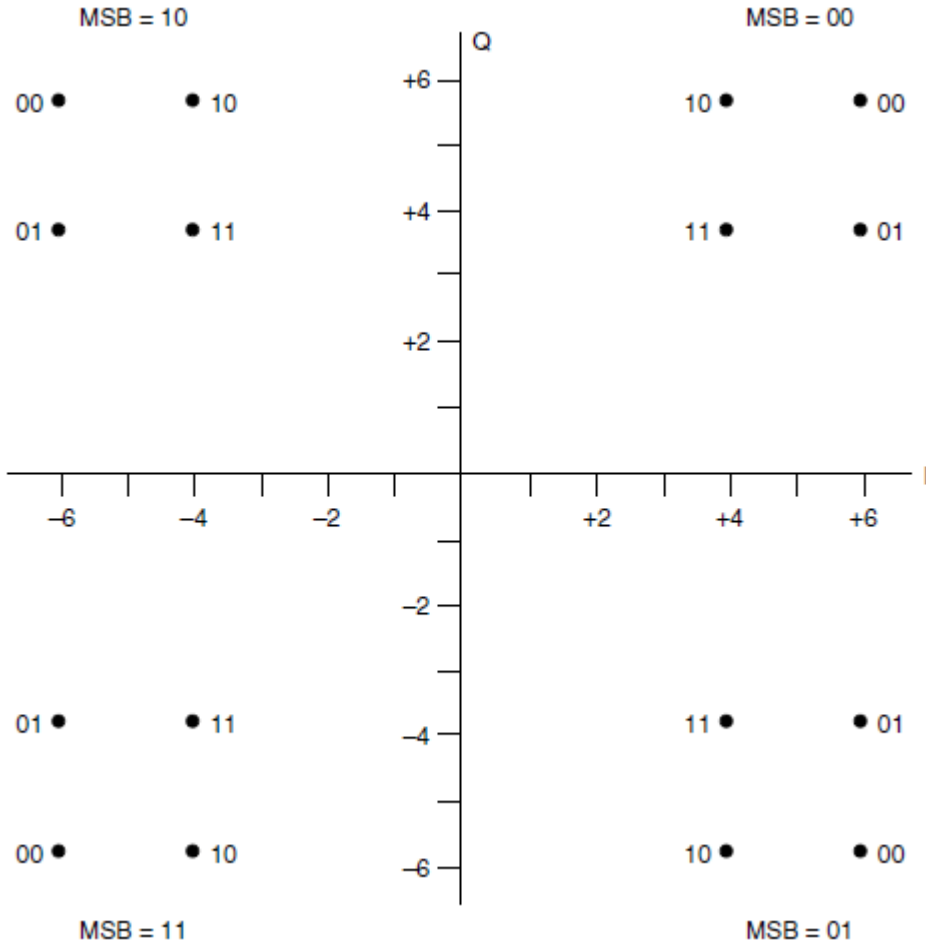
- Quadrature Amplitude Modulation (QAM)
  - Amplitude modulation on both quadrature carriers
  - $2n$  discrete levels, if  $n=2$  then same as QPSK
- Extensively used in microwave links
- DVB-T uses QAM
- 4-QAM



# 16-QAM

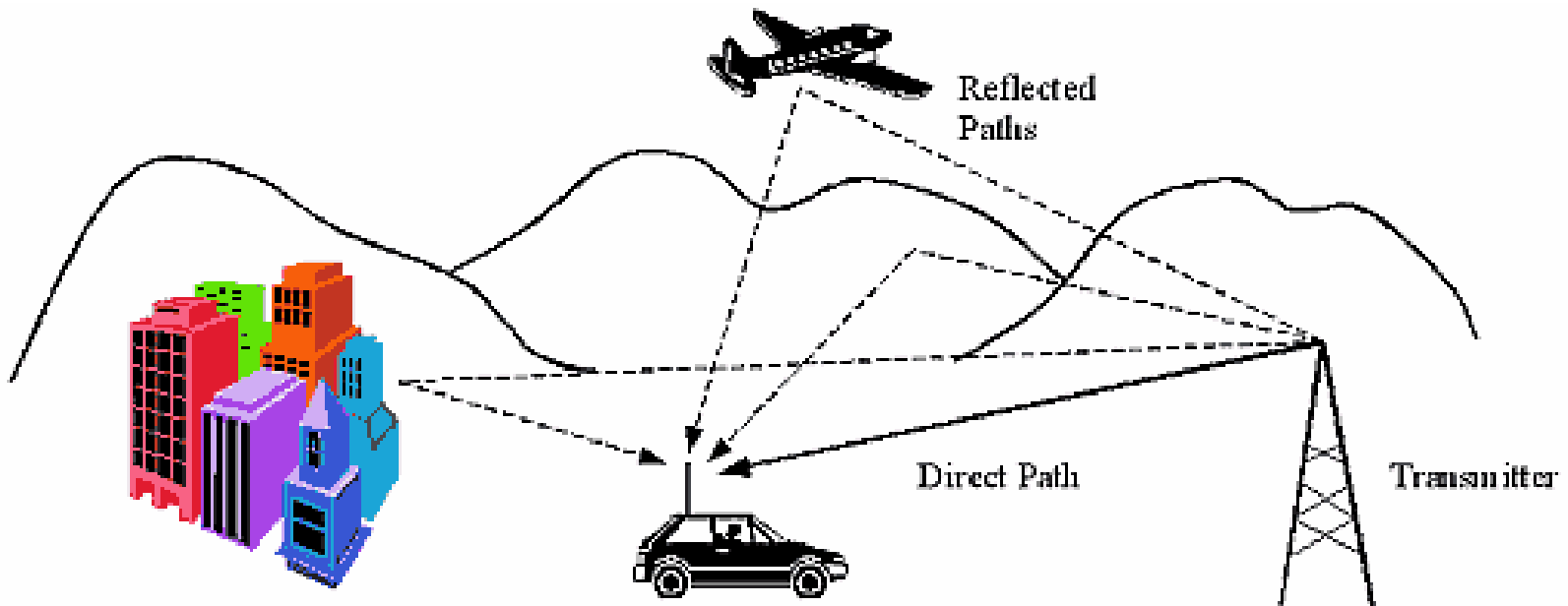


# 16-QAM



# OFDM

- Orthogonal Frequency Division Multiplex
- Satellite and Terrestrial
- Multipath signals
- Ghosting



# Inter-Symbol Interference

**Transmitted signal:**

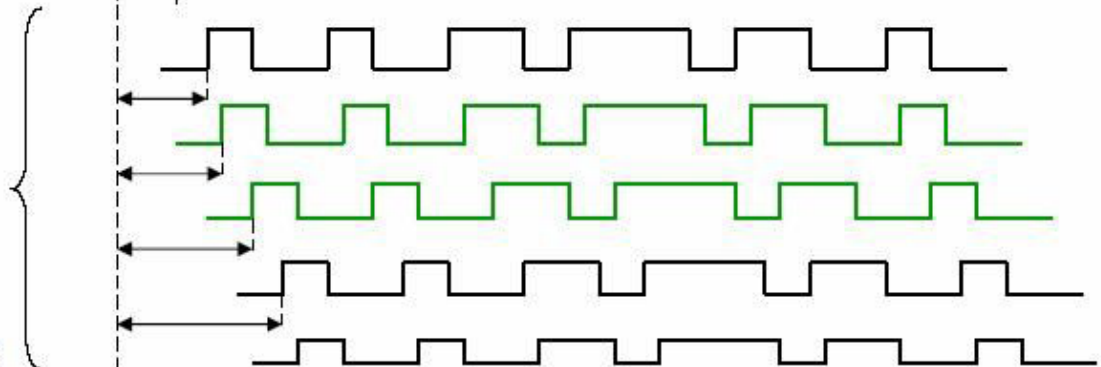


**Received Signals:**

Line-of-sight:



Reflected:



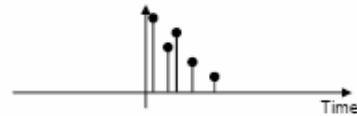
Delays

The **symbols add up**  
on the channel  
→ **Distortion!**



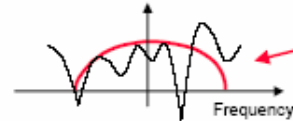
# Parallel Transmission

Channel impulse response



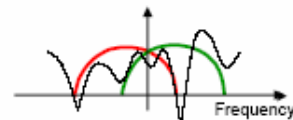
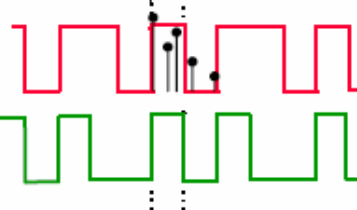
Channel transfer function

1 Channel (serial)

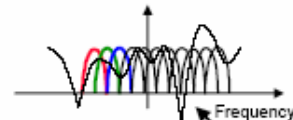
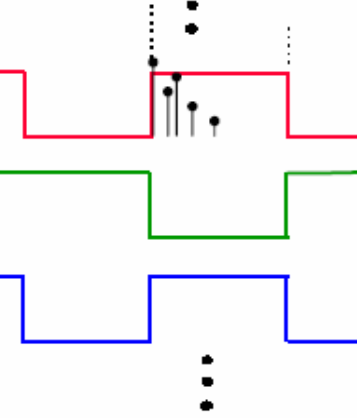


Signal is "broadband"

2 Channels

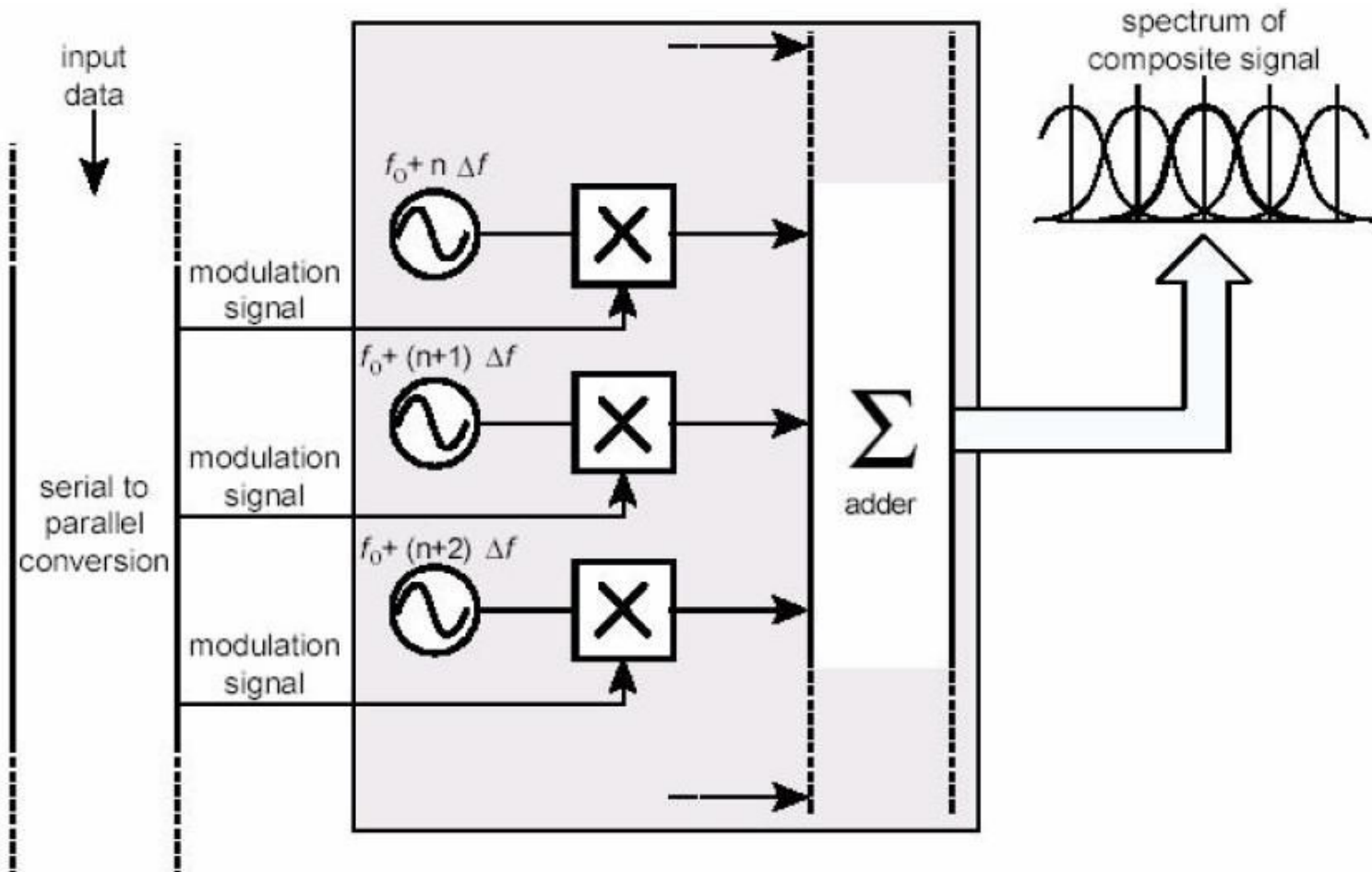


8 Channels



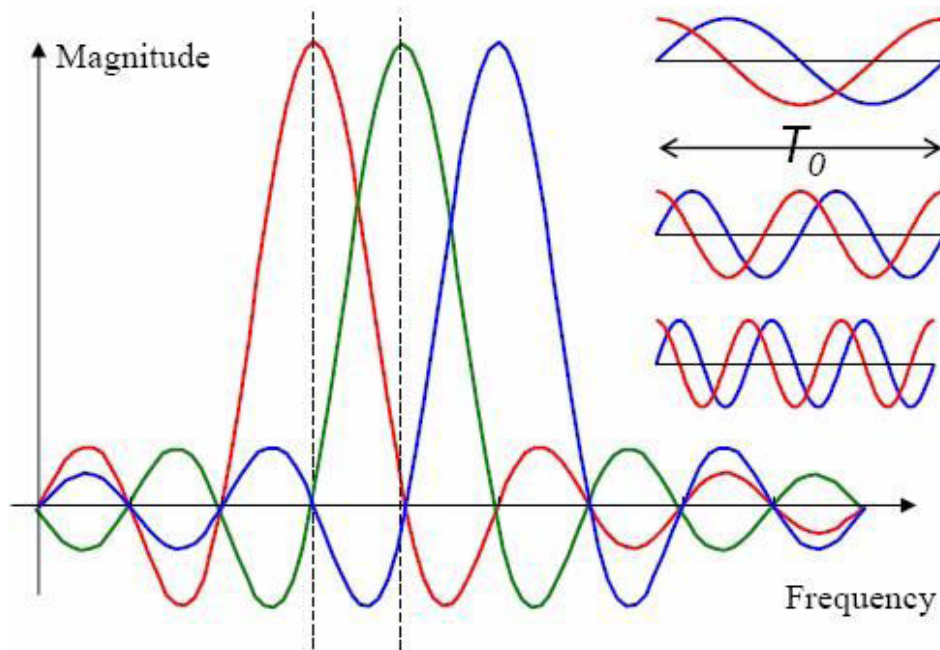
Channels are "narrowband"

# Visualization of OFDM



# Spectrum of the modulated data symbols

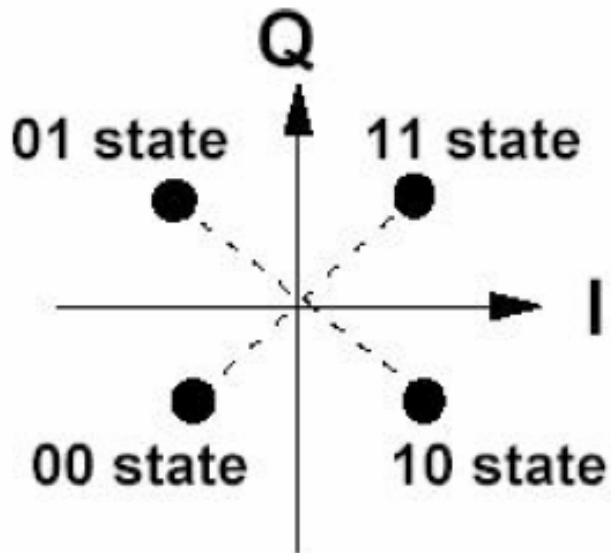
- Rectangular Window of duration  $T_0$
- Has a sinc-spectrum with zeros at  $1/T_0$ , other carriers are put in zeros
- Sub-carriers are orthogonal





# Carrier Modulation

- Symbol of sub-carrier is QPSK modulated
- Other symbol-alphabets can be used as well (BPSK, m-QAM)

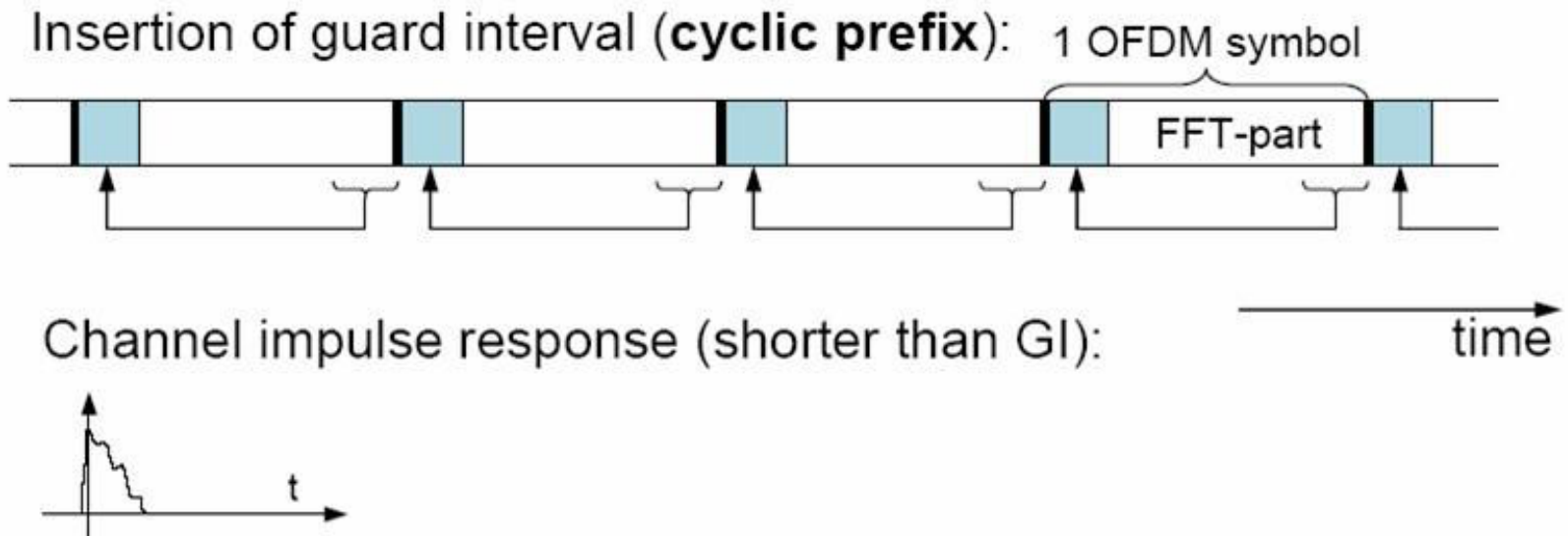


Phase of carrier:  
 $\pi/4, 3\pi/4, 5\pi/4, 7\pi/4$

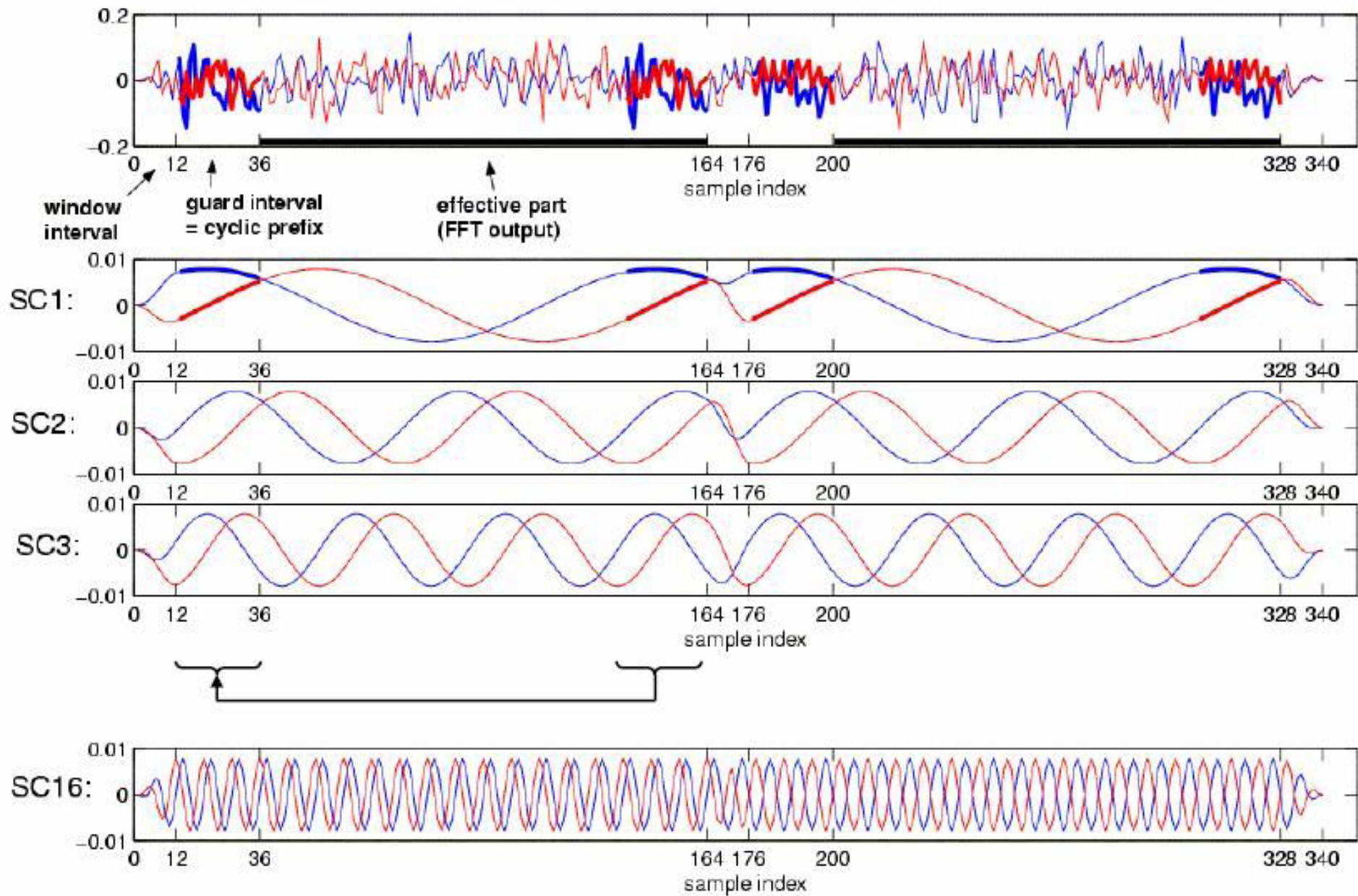
Output waveform is  
sum of modulated  $\pm$   
Cosine and  $\pm$ Sine wave

# Guard Interval - GI

- During the most of the symbol duration – signals in phase
- The only problem – transition due to the channel
- Channel response in time

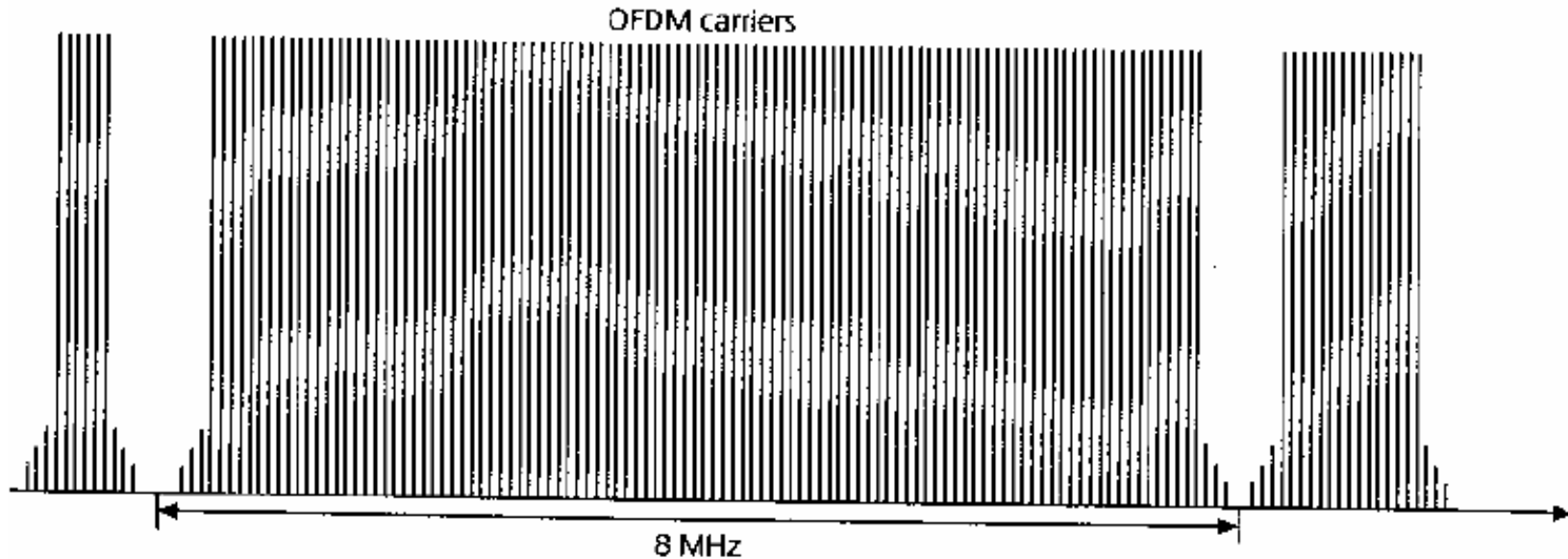


# time-domain OFDM signal:



# OFDM Spectrum

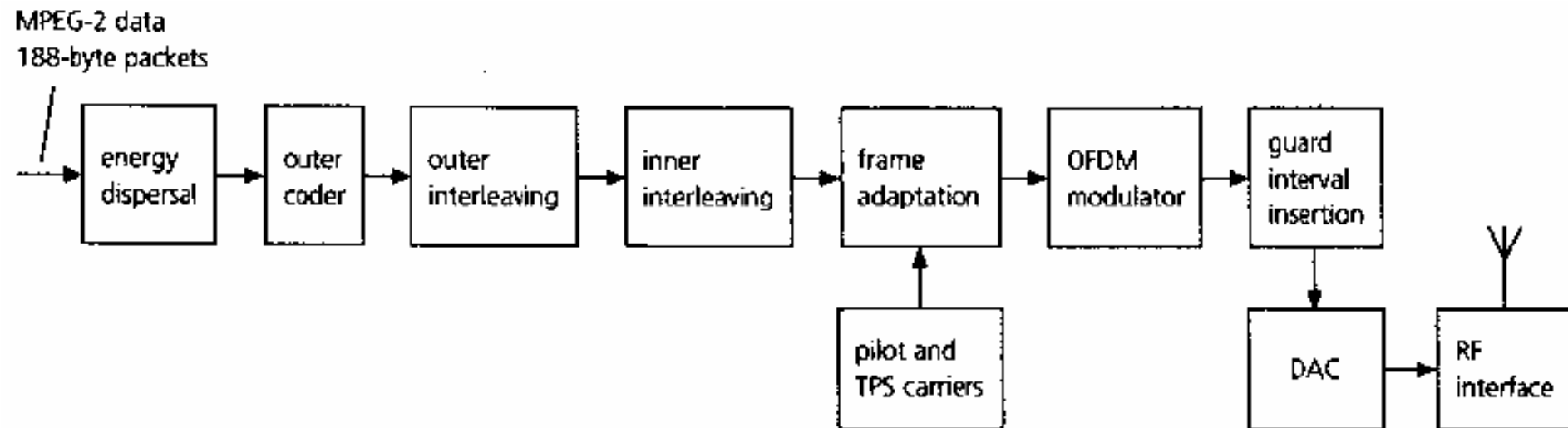
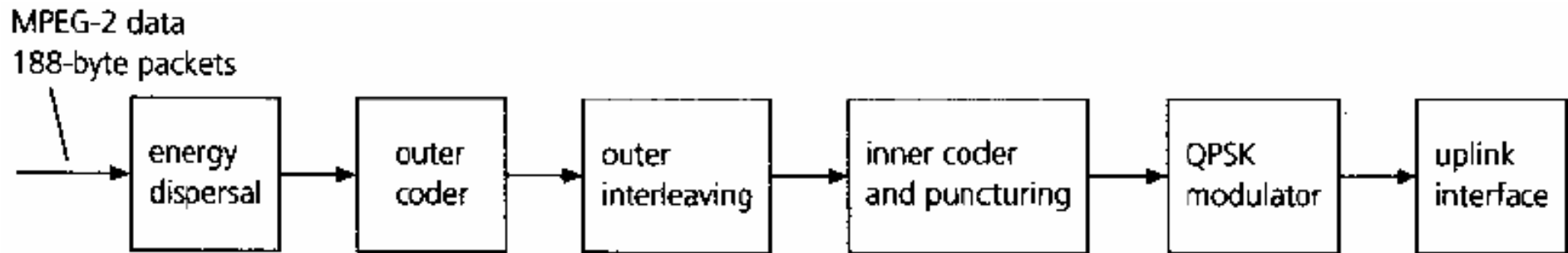
- Uniformly distributed – power efficiency
  - The same as a conventional TV channel – 8MHz
  - Same transmitter
  - High tolerance to fading and multipath interference
- Good for single channel frequency network SFN



# COFDM

- Includes powerful FEC – thus Coded
- Monitors the channel noise and narrow band interference
- Used in DVB-T
- 2k - 2048 carriers, 8k - 8192

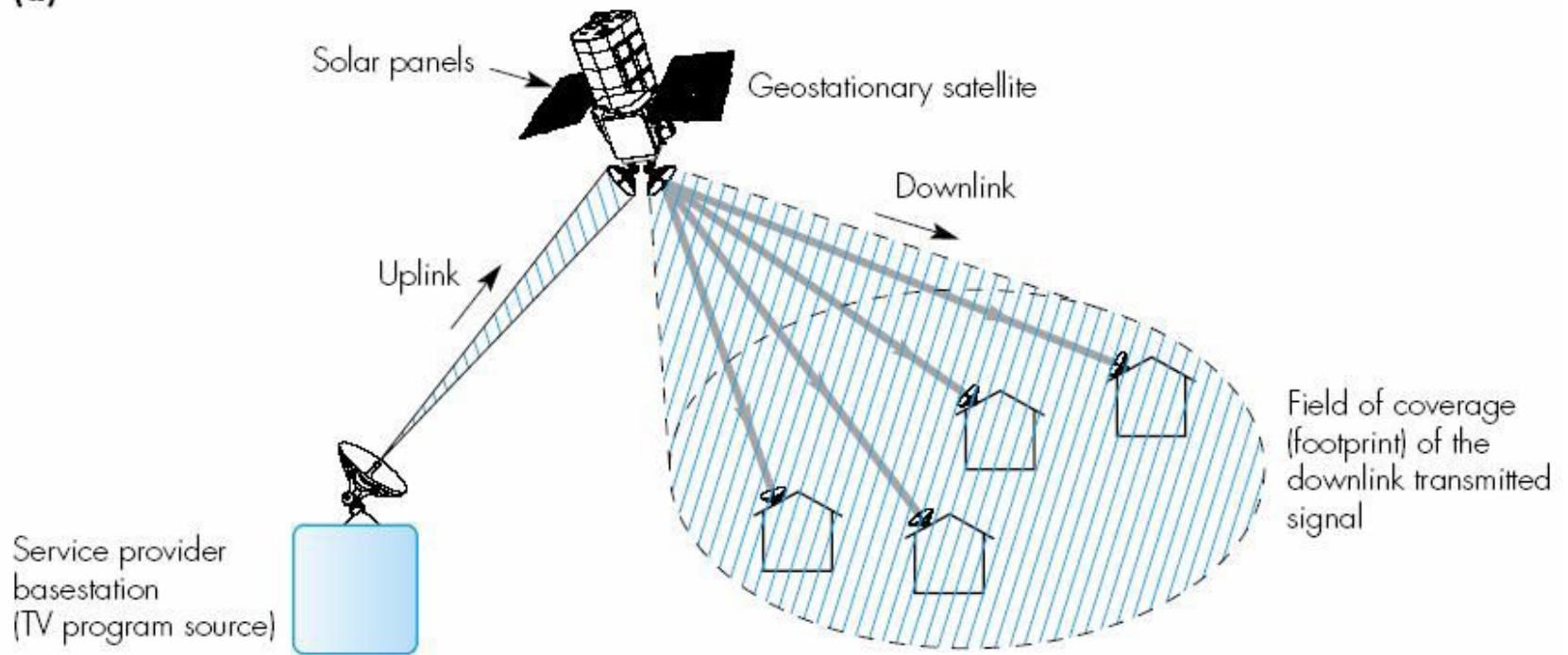
# DVB-S and DVB-T



# Satellite TV

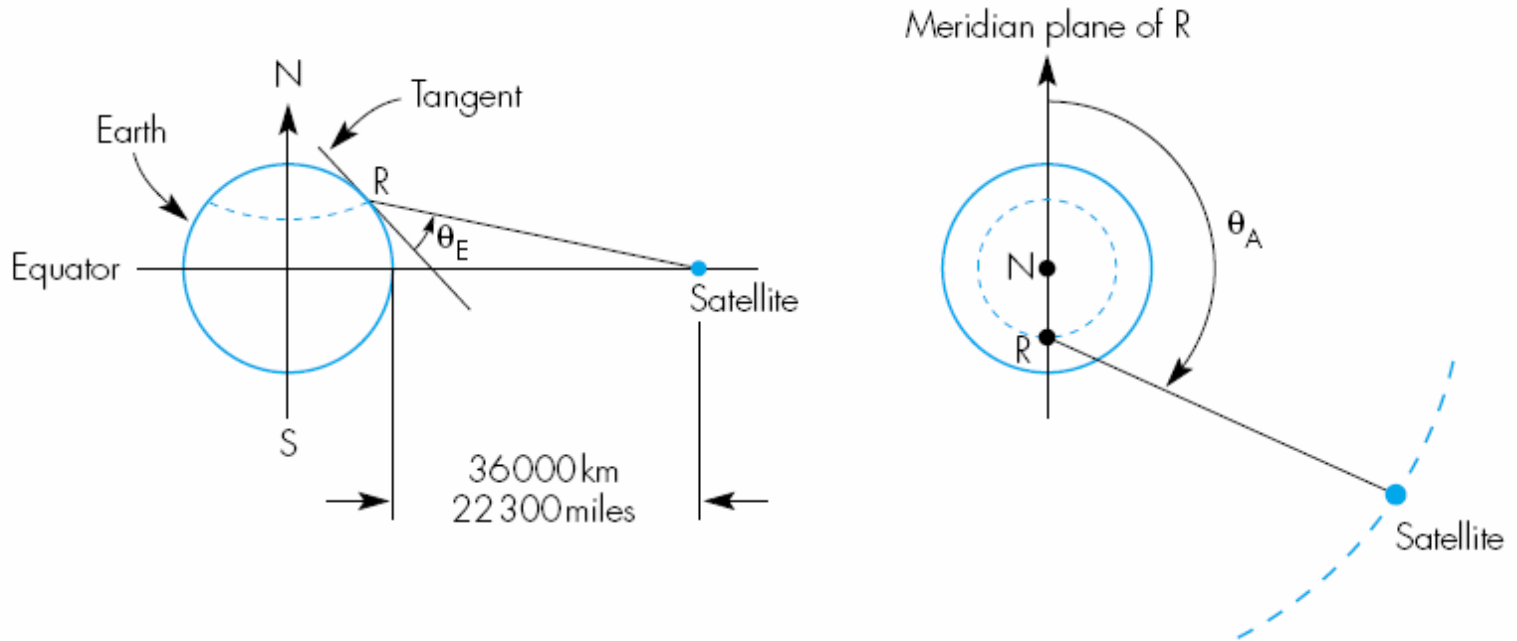
- *Geostationary satellites: satellites are geosynchronous*
- *36000km above equator*

(a)



# Satellite positioning

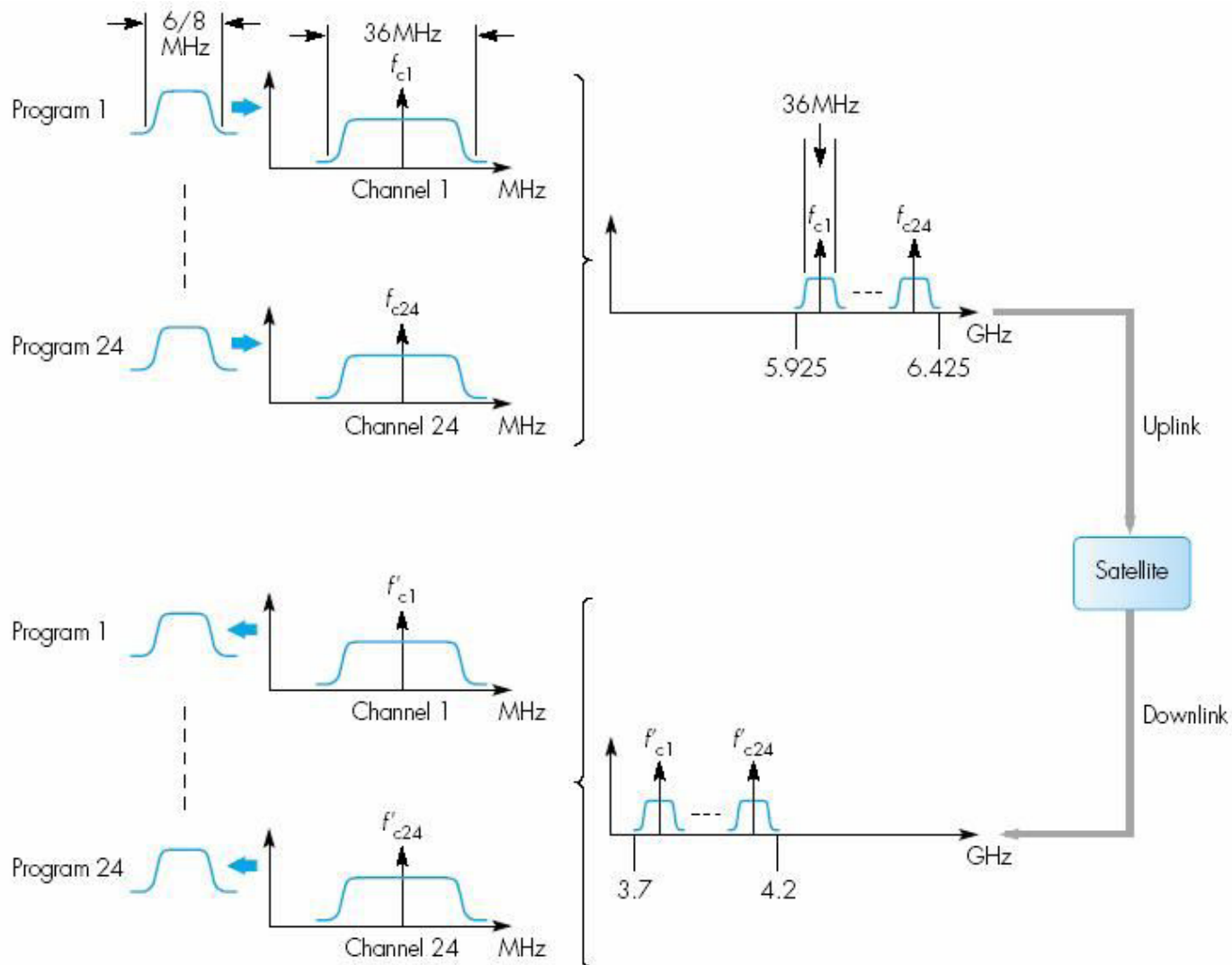
- *On-board motors make sure that the tolerance is  $0.2^\circ$*
- *Elevation and Azimuth define satellite's position*



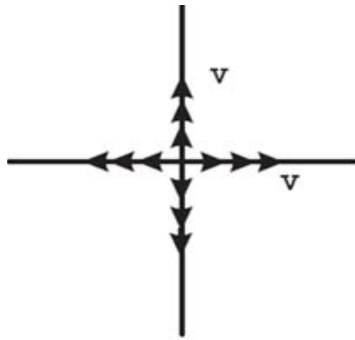
$\theta_E$  = angle of elevation       $\theta_A$  = azimuth  
(N = north, S = south, R = point of reception on earth's surface)



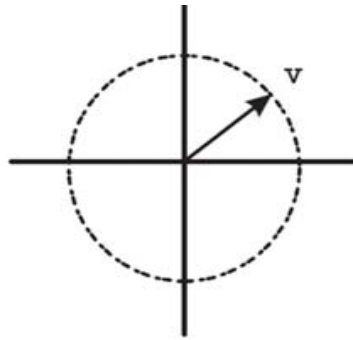
# Frequency allocation



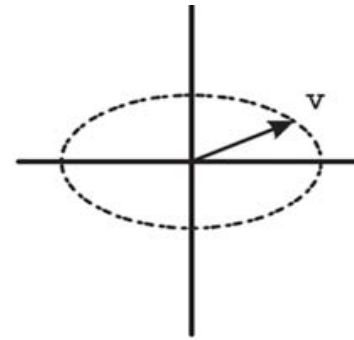
# Orthogonal polarization



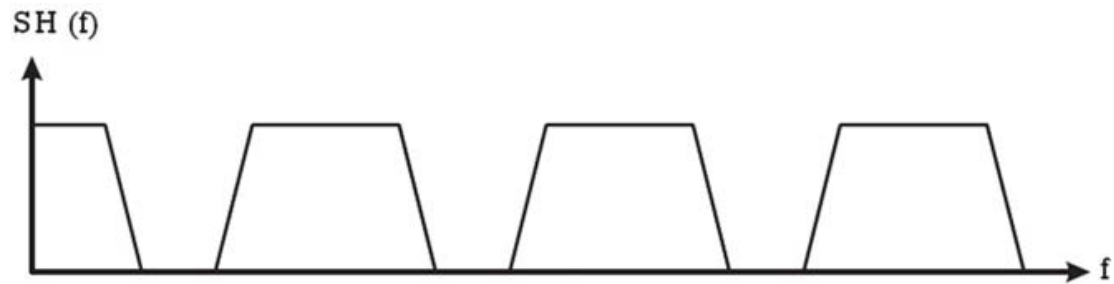
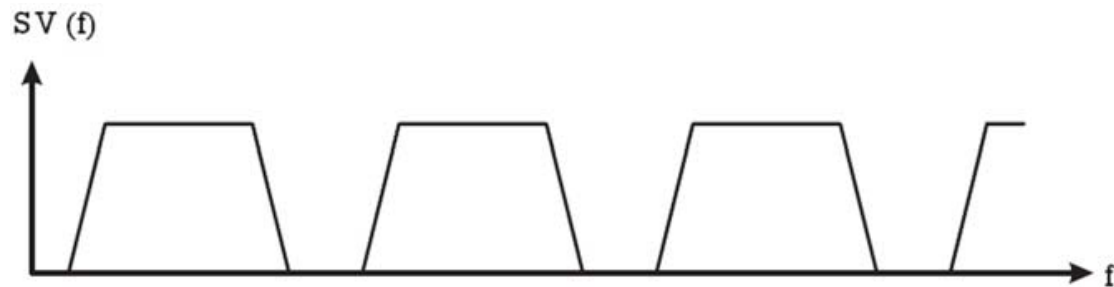
(a)



(b)

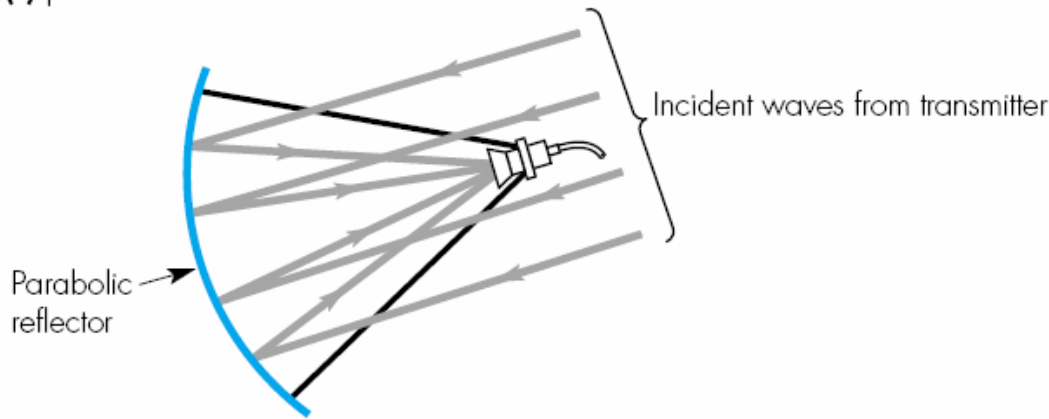


(c)



# Antenna designs

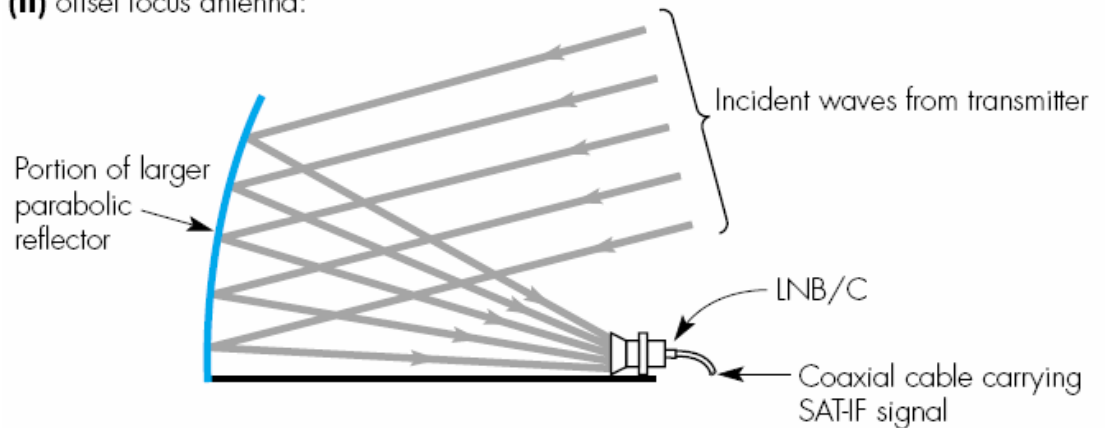
(i) prime focus antenna:



*Low-noise block converter – LNB/C*

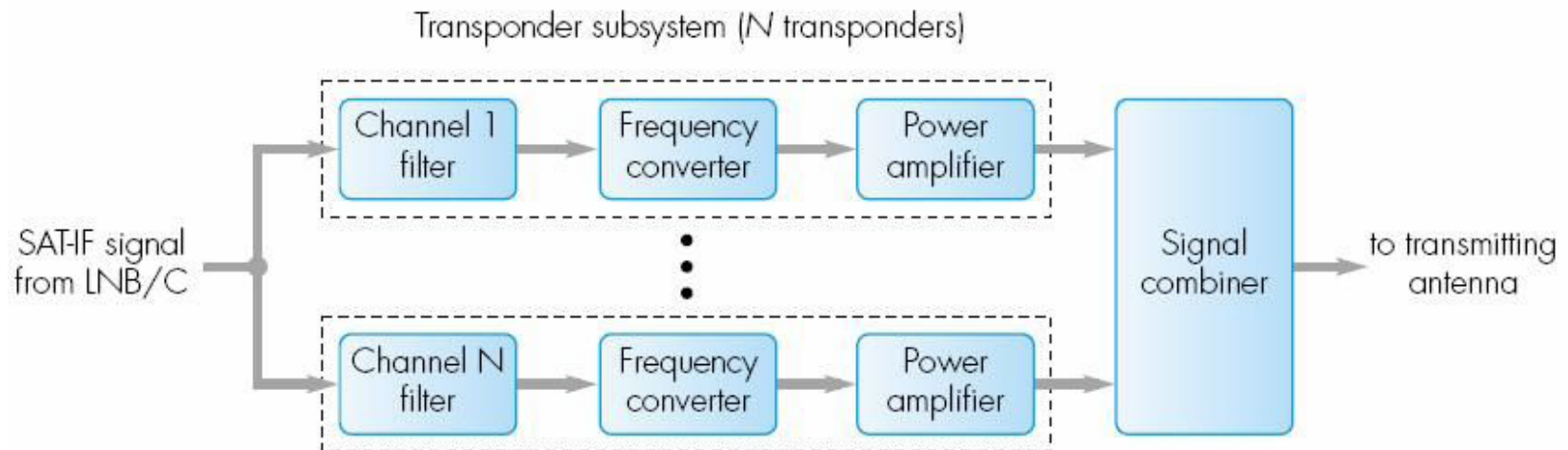
*Satellite intermediate frequency (SAT-IF)*

(ii) offset focus antenna:



# Transponder

- *Microwave power amplifiers*
- *Channels separated*
- *Combined signal is broadcasted to footprint*
- *At receiver side - inversion*



LNB/C = low-noise block/converter

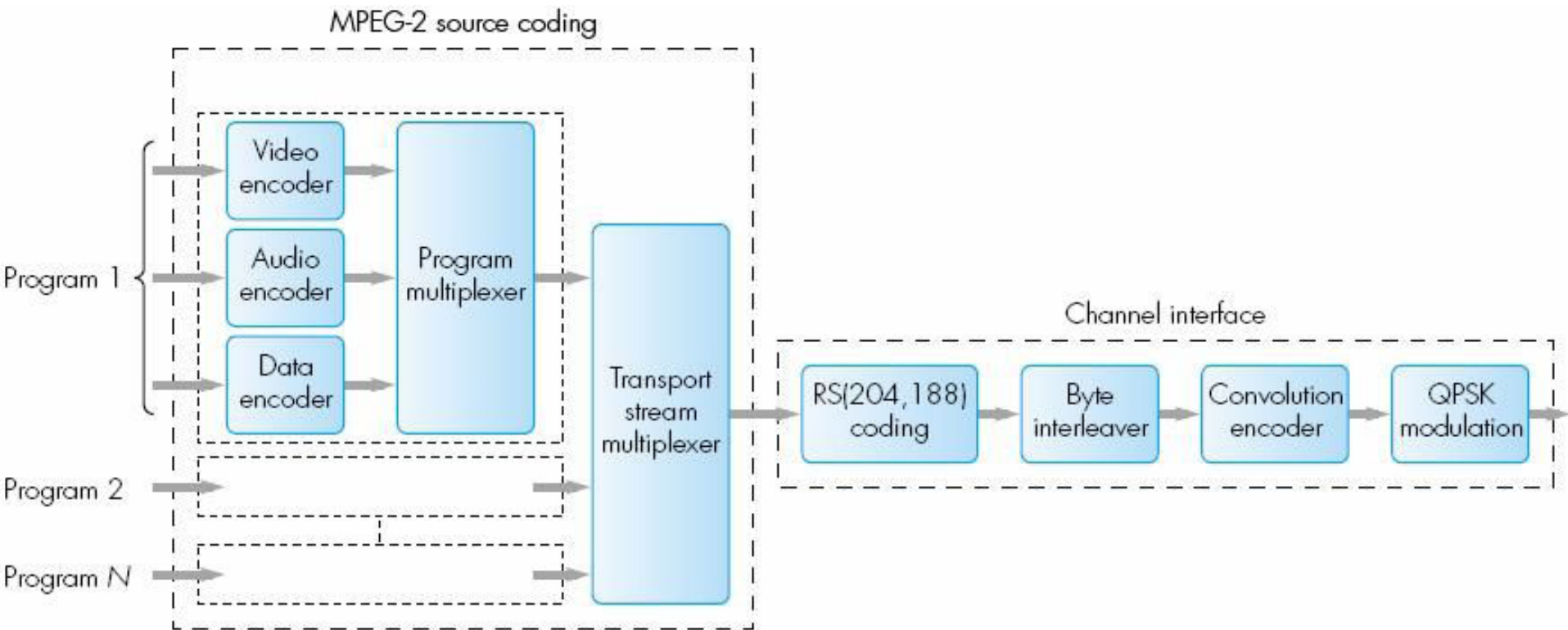
SATIF = satellite intermediate frequency

# Digital Television

- *QPSK = 4-QAM, 2b/symbol*
- *DTV KU band – 10.7 – 14.5 GHz*
- *Downlink*
  - *10.7-11.7GHz analogue TV*
  - *12.2-12.7GHz digital TV USA, 11.7-12.5GHz EU*
- *DBS, USA, 32x24MHz – 40Mbps*
- *DVB-S, EU, 40x33MHz – 55Mbps*

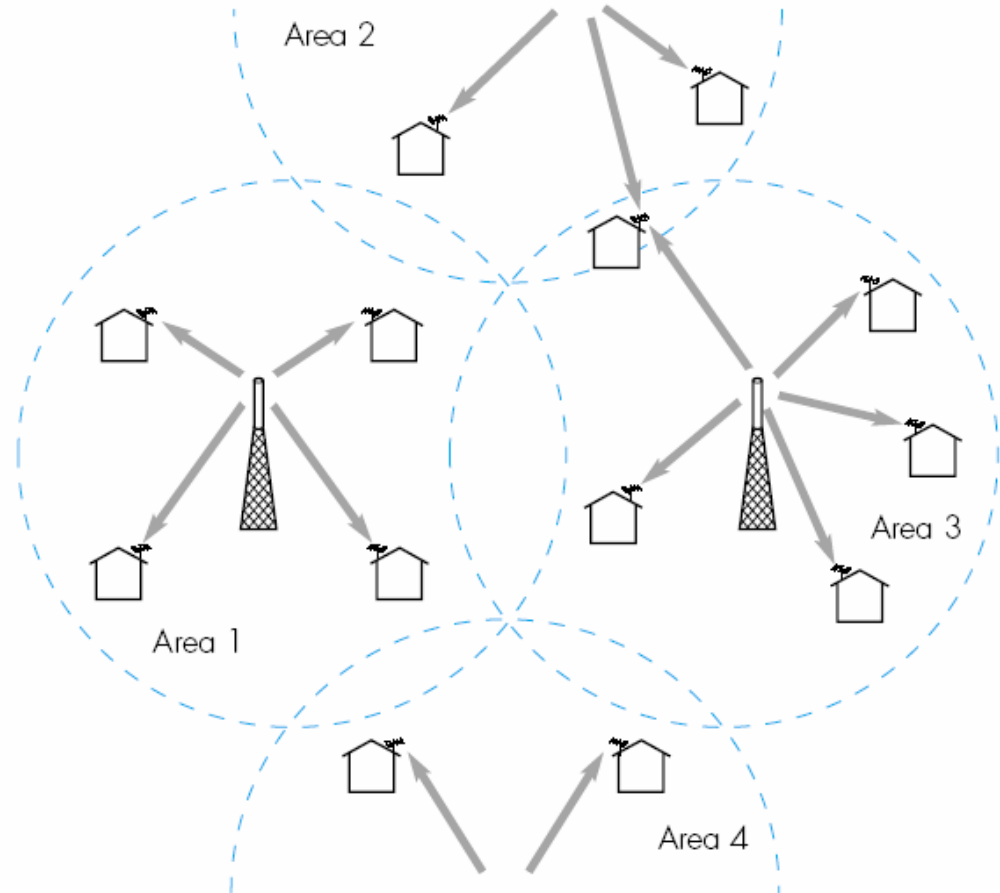
# SATV Channel Interface

- *MPEG-2 Transport stream multiplex*



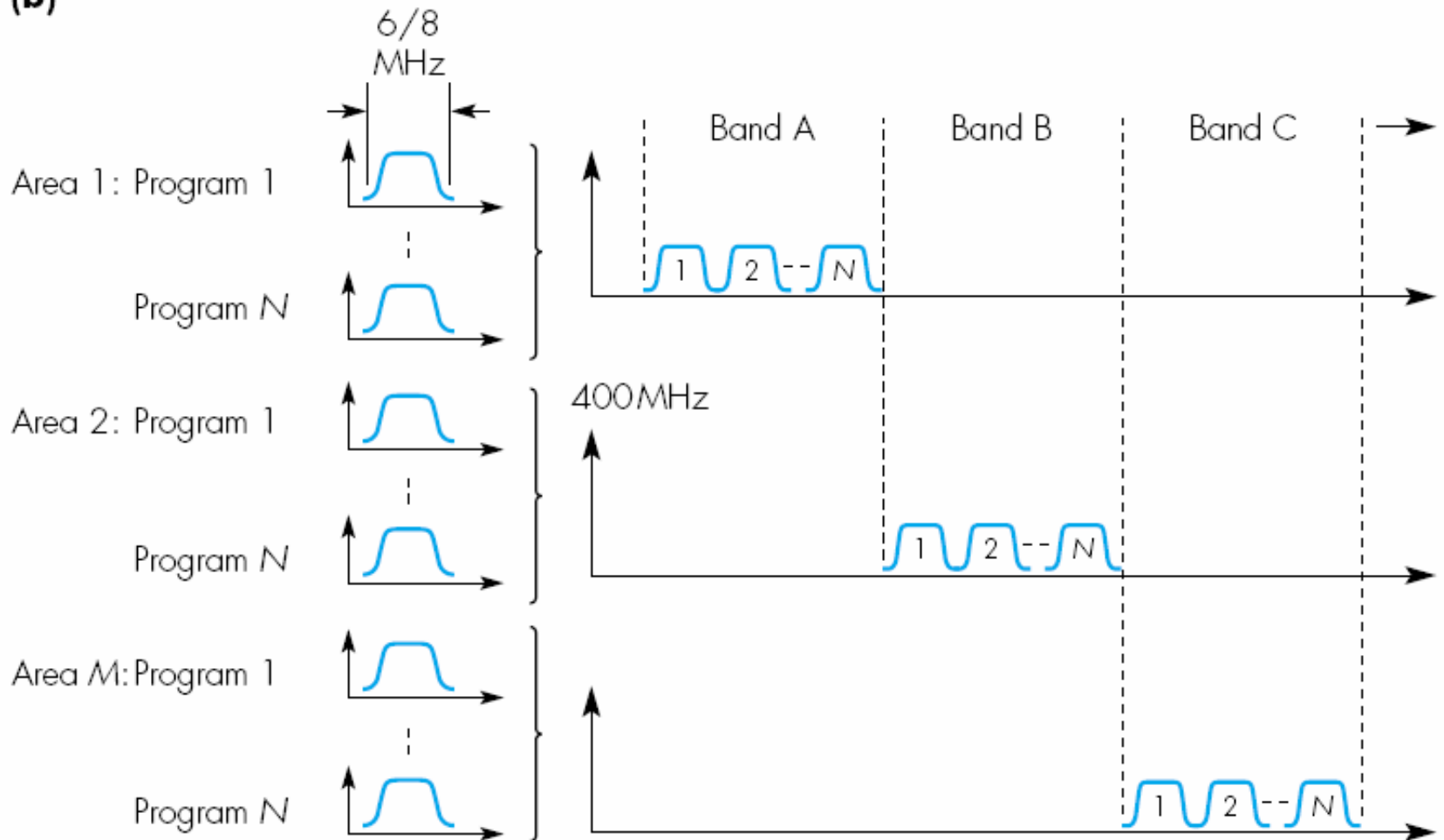
# Terrestrial Television

- *VHF*
- *UHF*
- *47-860MHz*
- *>100MHz – linear*
- *More transmitters*
- *MFN - depicted*



# MFN Frequency Usage

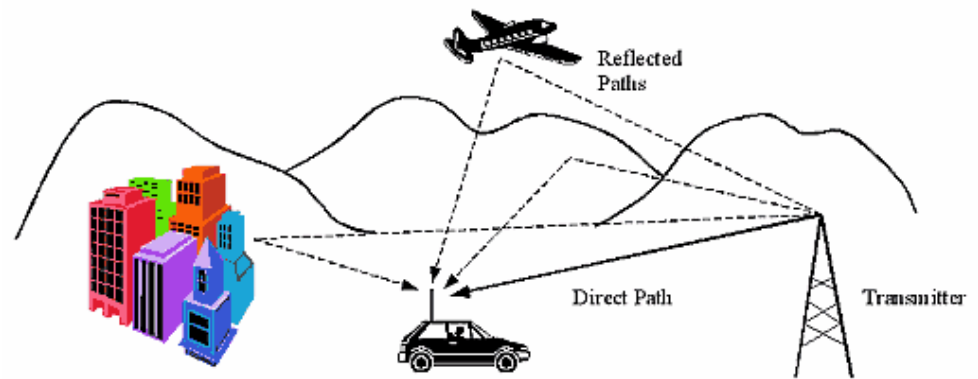
(b)





# Digital Terrestrial TV

- *Different modulation*
- *Microwaves reflect from objects*
- *Multipath transmission*
- $\lambda = c/f = 3 \times 10^8 / 500 \times 10^6 = 0.6 \text{ m}$
- *Reflection, refraction*
- *Multipath dispersion, delay spread*
- *Inter-symbol interference*
- *Solution COFDM – coded OFDM*



# Inter-Symbol Interference

**Transmitted signal:**

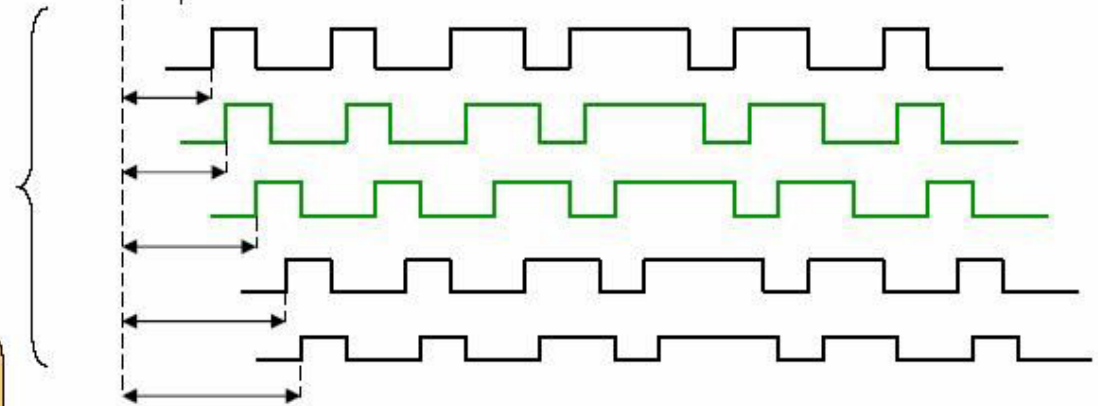


**Received Signals:**

Line-of-sight:



Reflected:



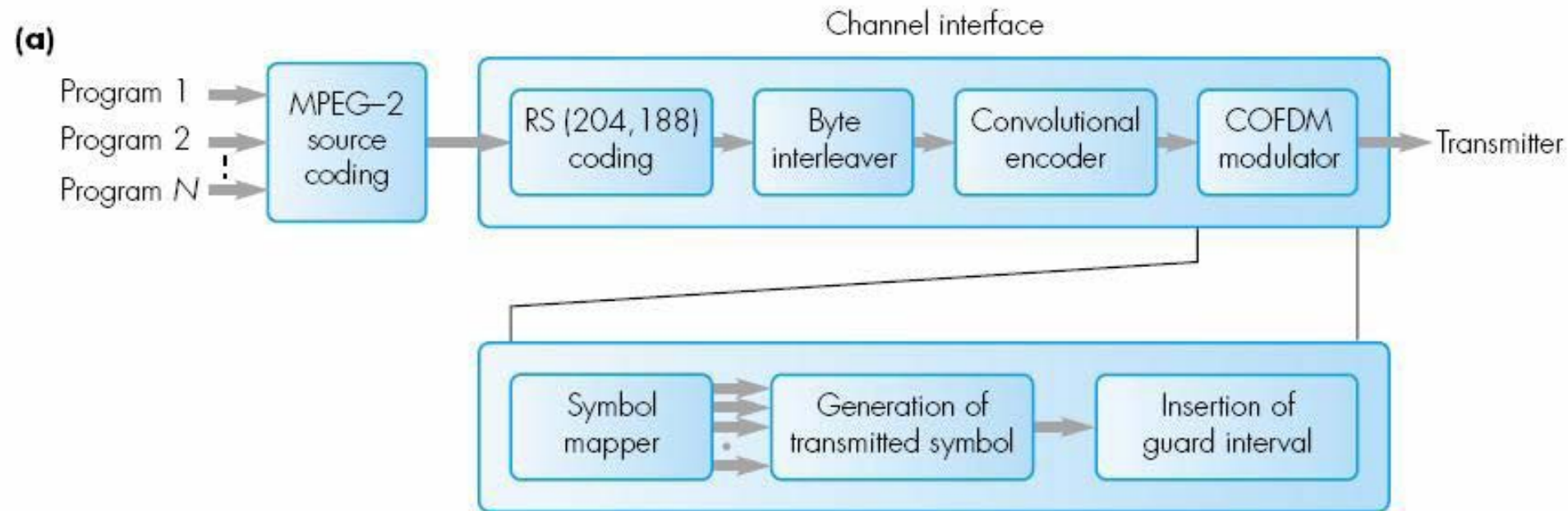
The **symbols add up**  
on the channel  
→ **Distortion!**

Delays



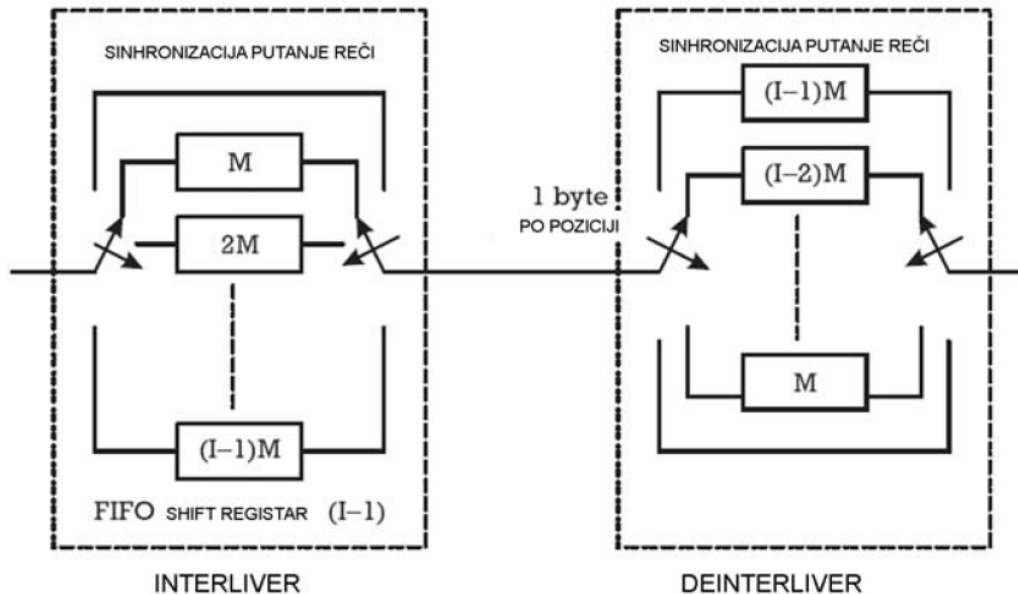
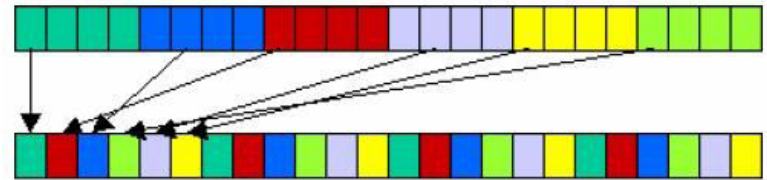
# Channel Interface

- $RS(204,188)$
- *Interleaving* – breaking down very long error bursts for FEC
- *By rearranging the order of transmission of the bytes*



# Forney interleaving

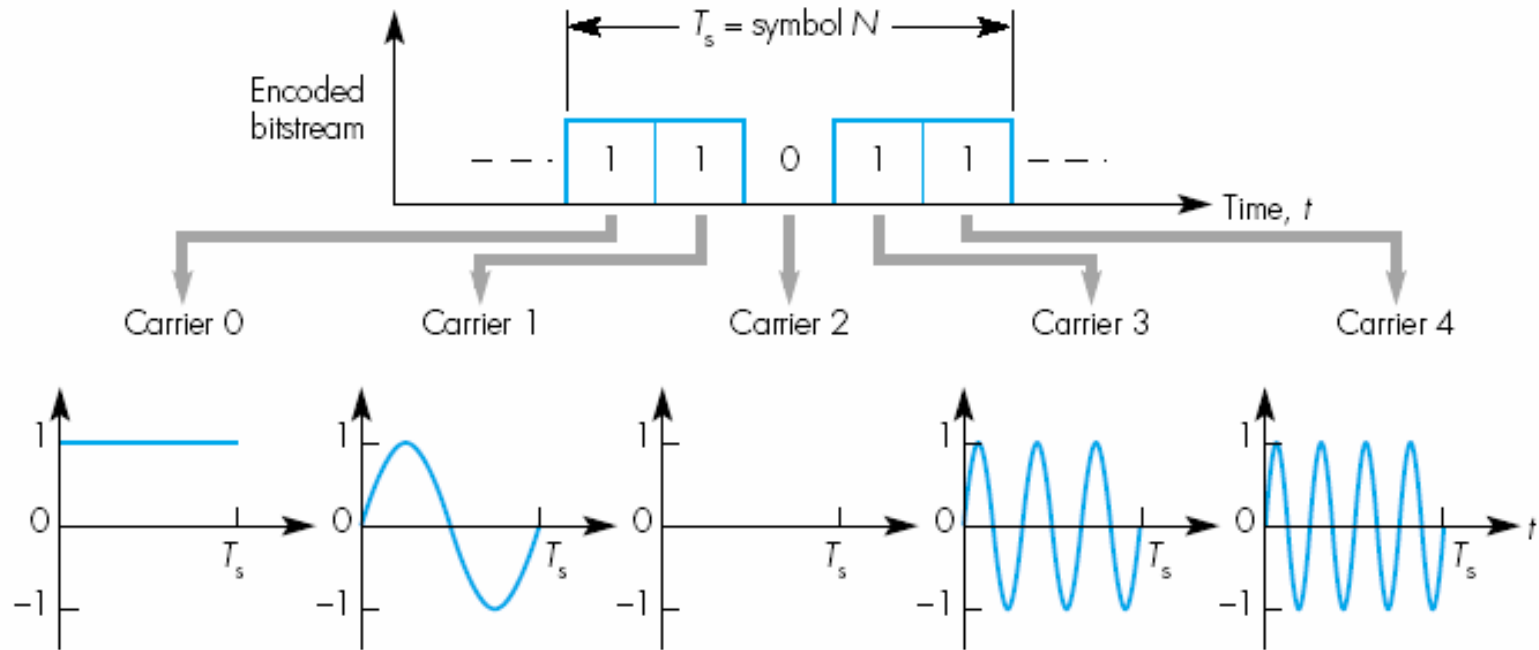
- To spread out the error bits – enables more efficient outer coding
- DVB-T:  $i=12$  branches  $L=204$  packet length,  
 $M=(L/i) \times j$  - FIFO buffer size,  $j=0, 11$



# COFDM Principles

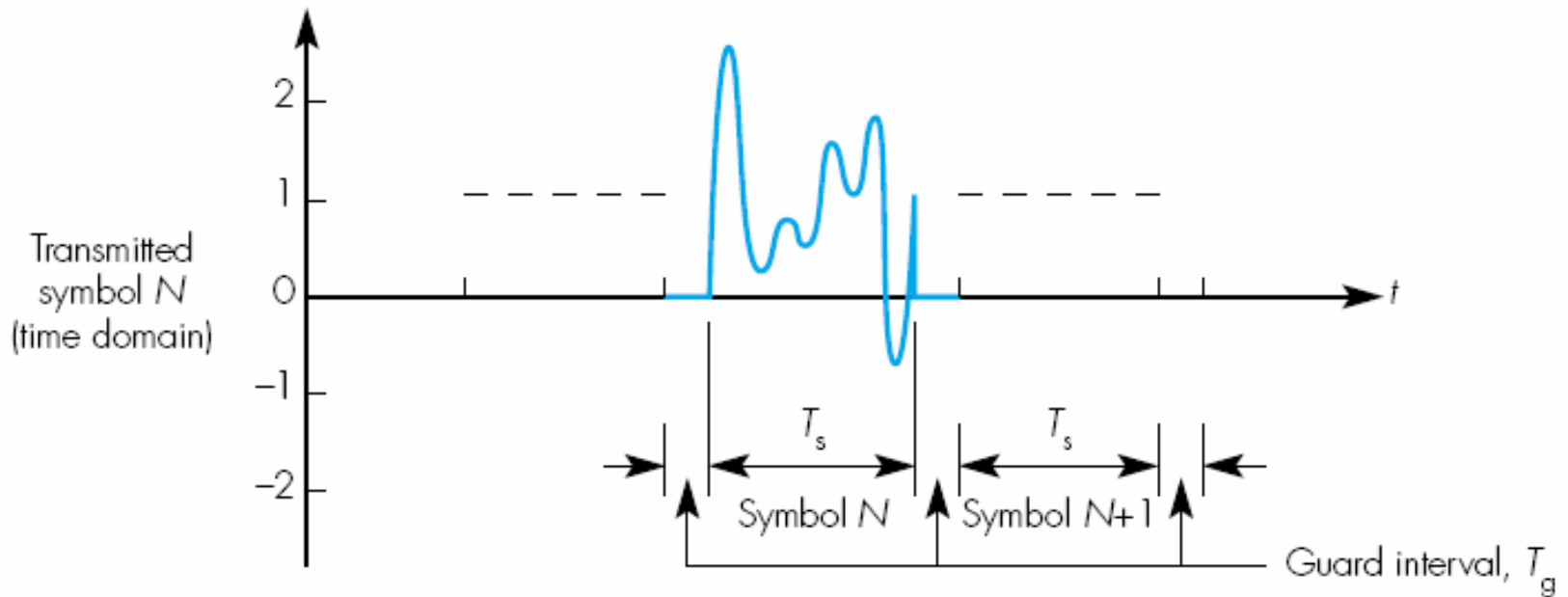
- Example of multiple orthogonal carrier modulation
- 5 carriers, ASK on-off keying, frequencies  $f_s$ ,  $2f_s$ ,  $3f_s$ , and  $4f_s$
- In the real DVB-T case, QPSK,  $N=[2k, 8k]$  carriers

(b)



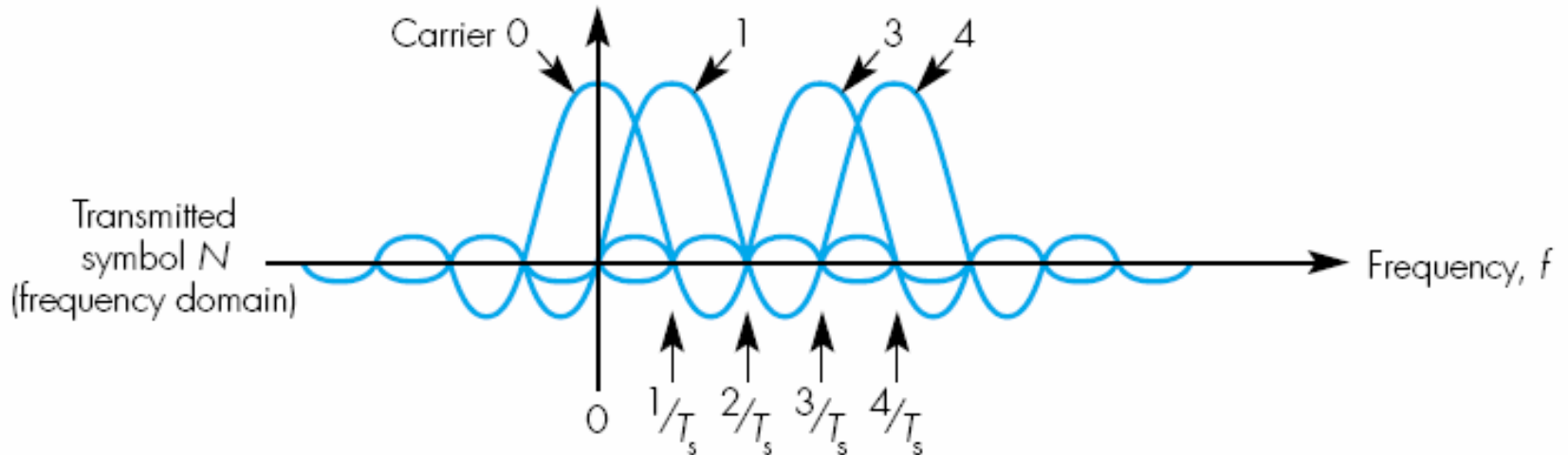
# Guard interval

- Receiver waits  $T_g$  (guard interval) before starting to process the symbol
- In order to make sure all delayed versions of the symbol have arrived



# Orthogonal carriers

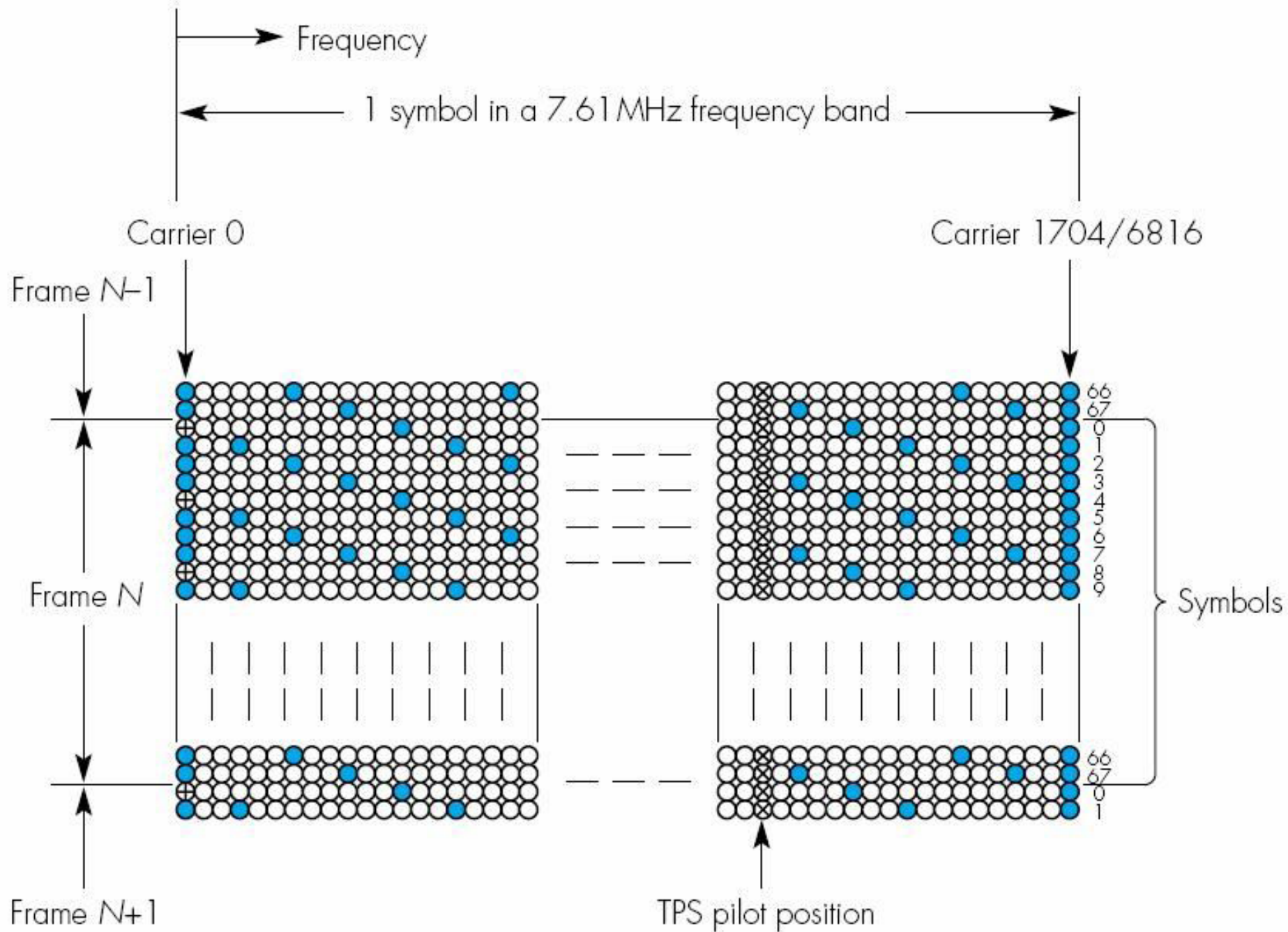
- *To be able to determine which subset of the carrier-ensemble is present in a given symbol, carriers need to be ORTHOGONAL*
- *At carrier frequency ALL other carriers are ZERO*
- *Strict fixed spacing of  $1/T_s$  between adjacent carriers*



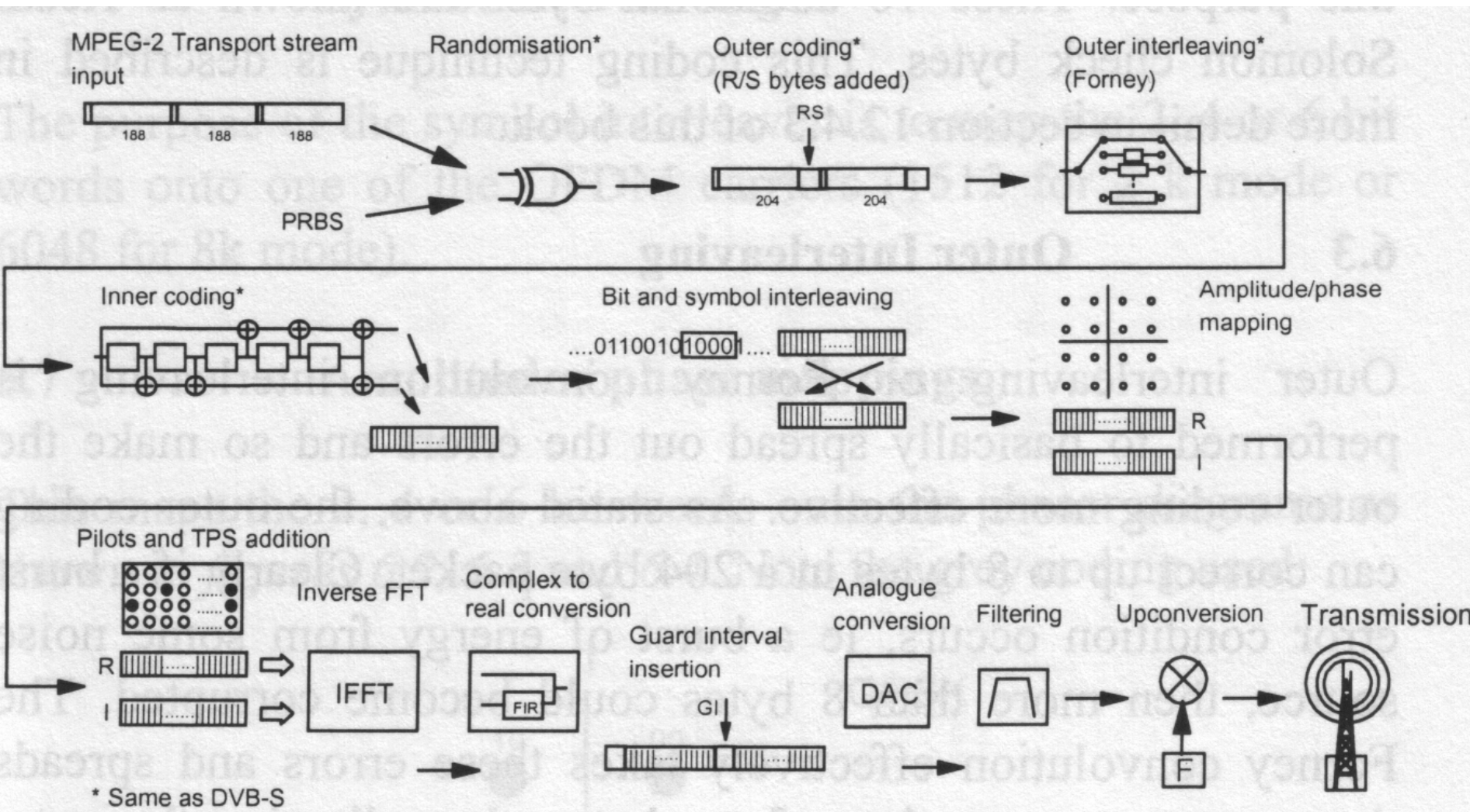
# Receiver Synchronisation

- *Synchronising symbols essential*
- *Stream of symbols is divided into 68-symbol FRAMES*
- *Number of active carriers 2k – 1705, 8k – 6817*
- *Each carrier modulated with*  
*4-QAM 16-QAM or 64-QAM*  
*2, 4 or 6 bits/signal transition*
- *Continual pilot*
- *Scattered pilot*
- *TPS pilot – information on operational parameters*





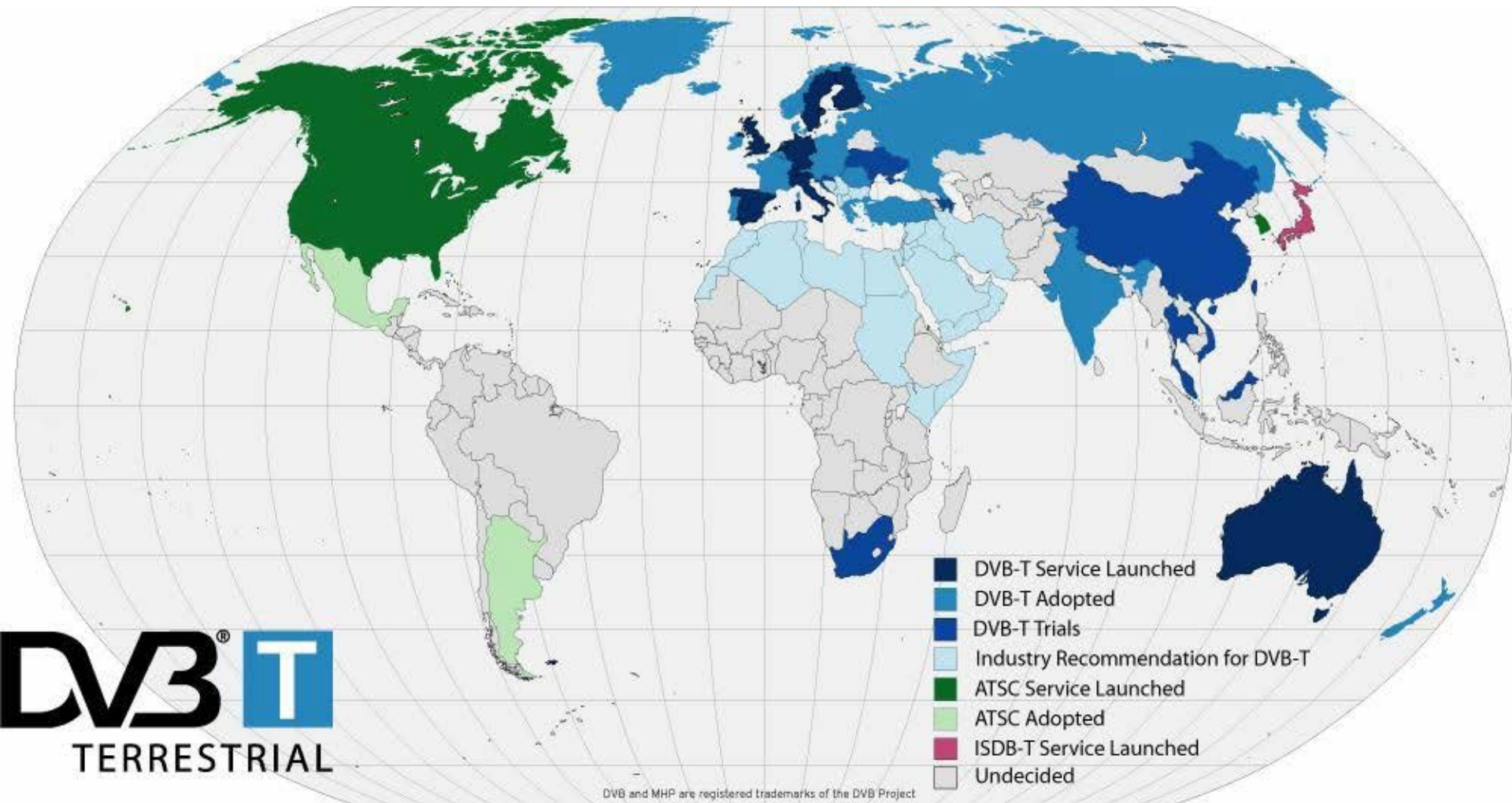
# DVB-T Transmission System



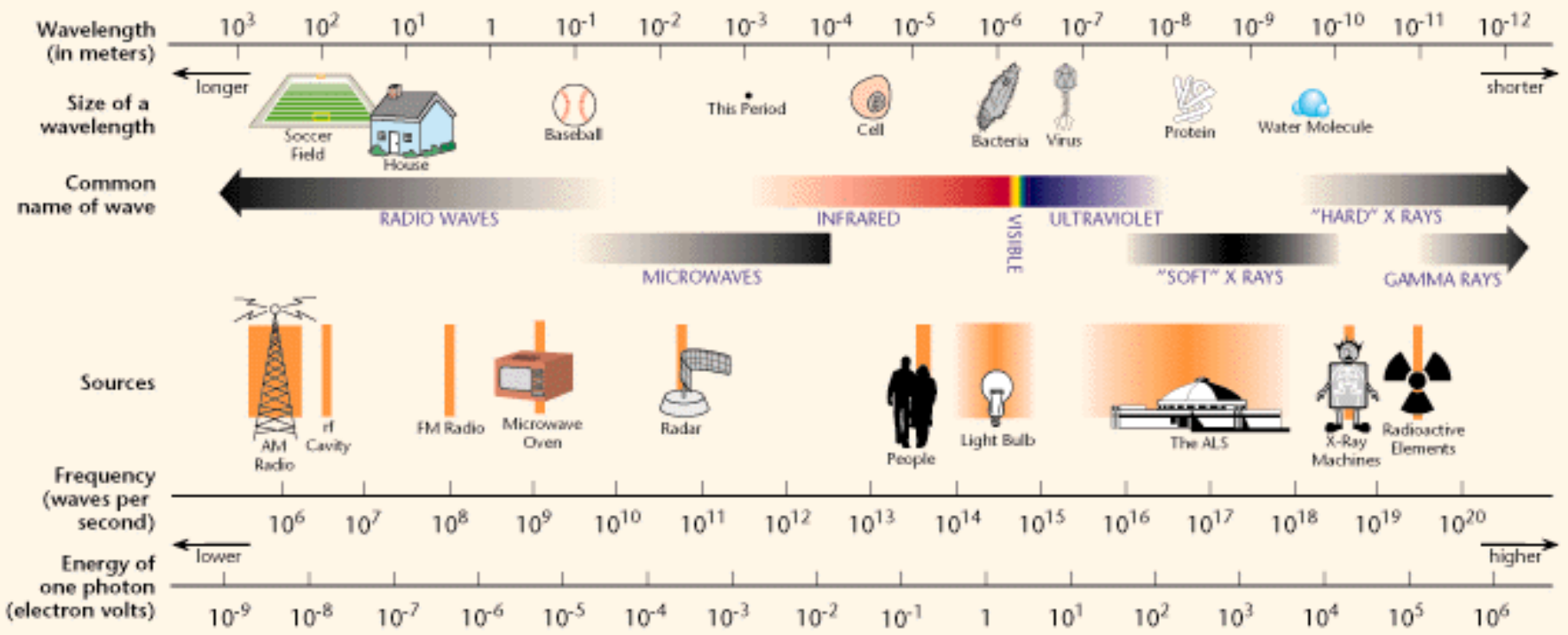
# DVB-T Bandwidth

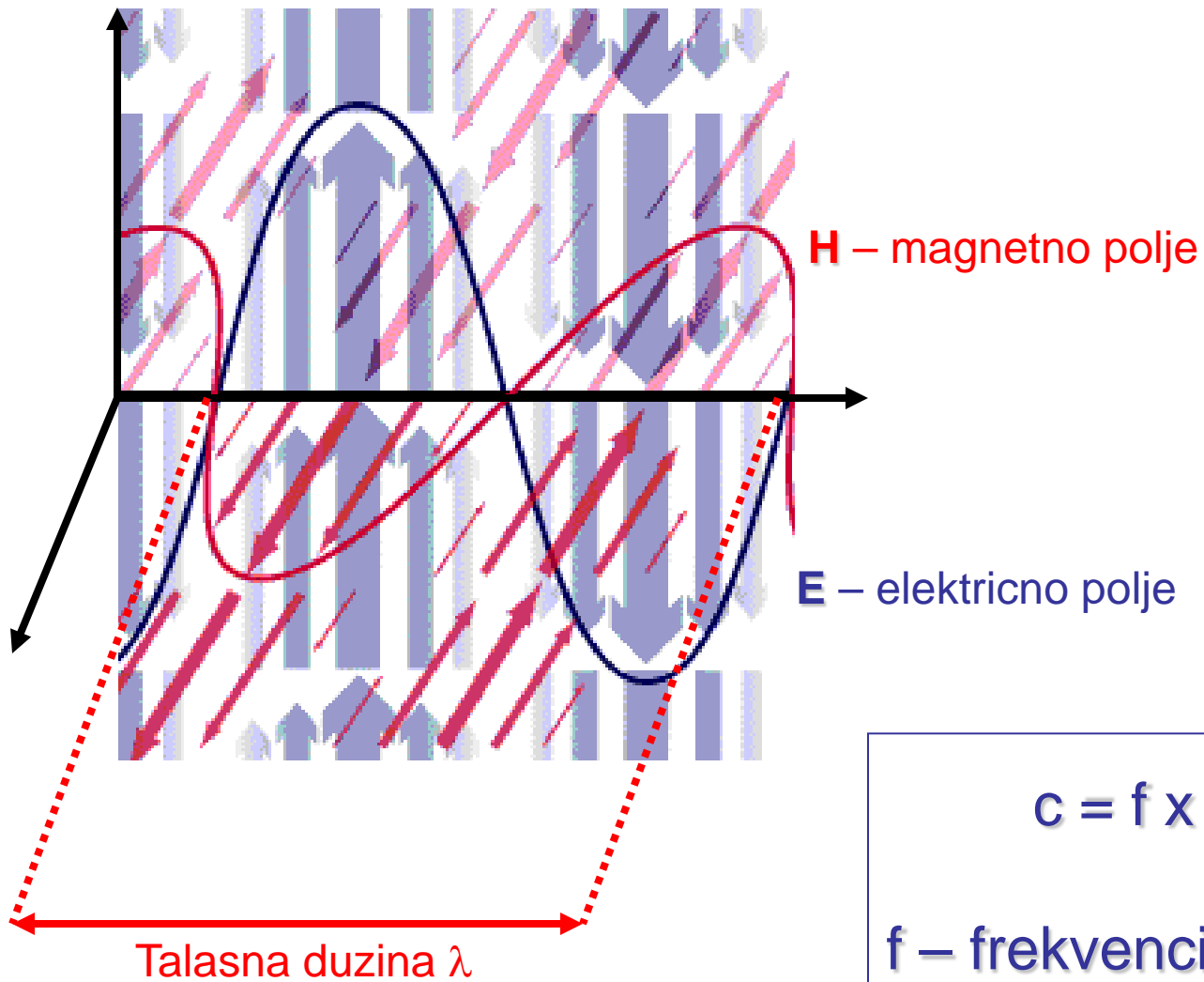
- *6/8 Mhz broadcast channels*
- *For 2k 16-QAM modulated carriers, removing pilots, guard, etc. the bandwidth is 24Mbps*
- *FEC and convolutional coder reduce this to 16Mbps*
- *This can be used to carry 4x4Mbps TV programmes in each 8MHz broadcast channel*

# DVB-T Coverage



# THE ELECTROMAGNETIC SPECTRUM





$$c = f \times \lambda$$

$f$  – frekvencija

$c$  – brzina svetlosti

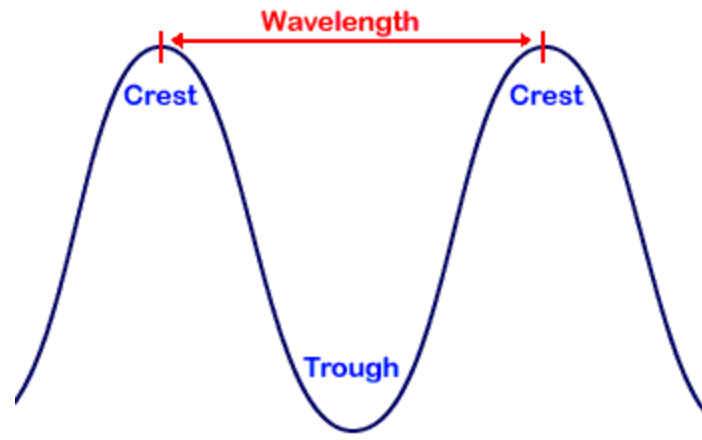
$\lambda$  - talasna duzina



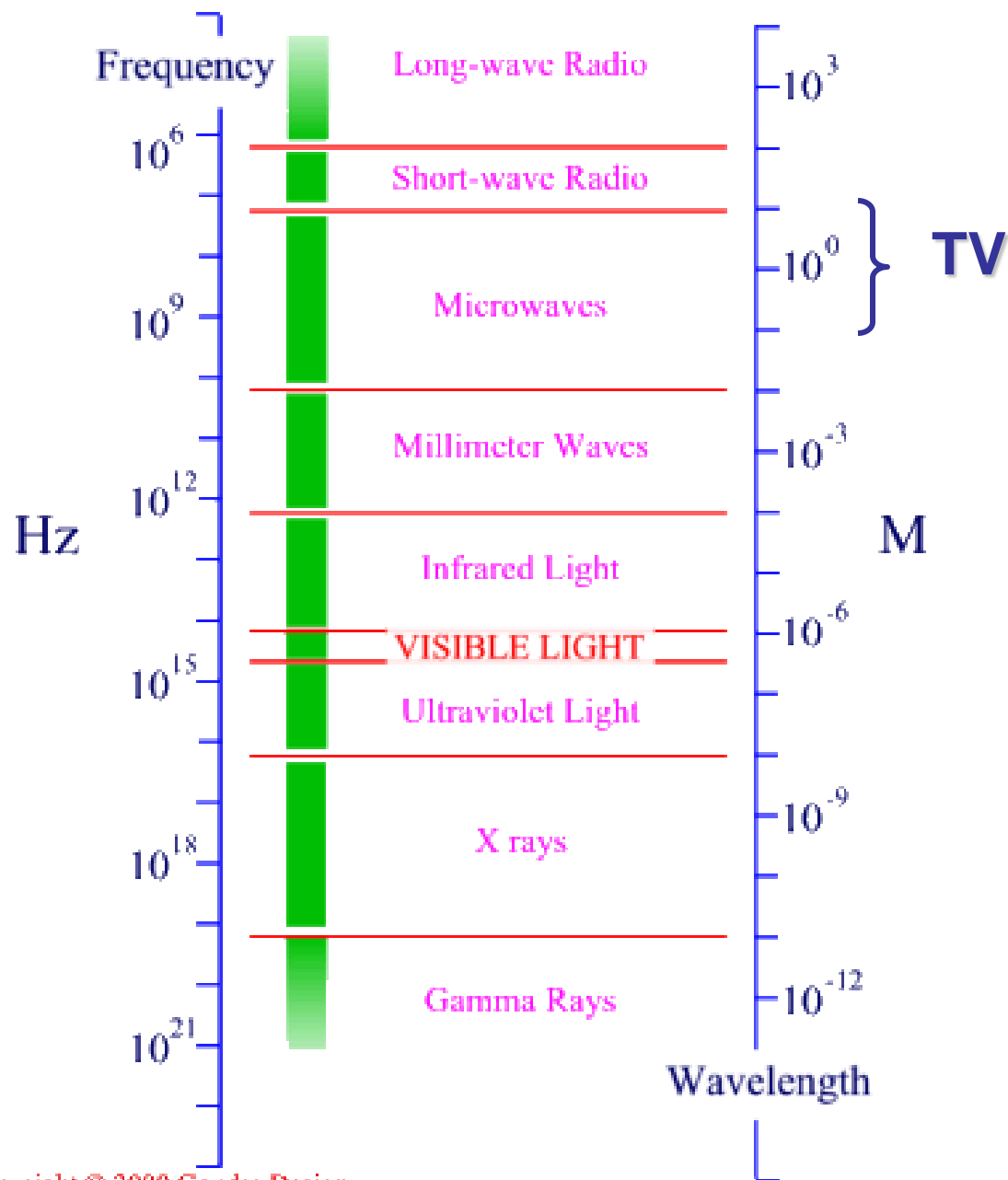
Wavelength in centimeters



About the size of...



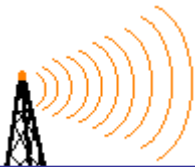
# The Electromagnetic Spectrum





## Radio Waves

$10^4 - 10^{-2}$  m /  $10^4 - 10^{10}$  Hz



ultra-low frequency (ULF)

3 - 30 Hz

extremely low frequency (ELF)

30 - 300 Hz

voice frequencies (VF)

300 Hz - 3 kHz

very low frequency (VLF)

3 kHz - 30 kHz

low frequency (LF)

30 kHz - 300 kHz

medium frequency (MF)

300 kHz - 3 MHz

high frequency (HF)

3 MHz - 30 MHz

very high frequency (VHF)

30 MHz - 300 MHz

**ultra high frequency (UHF)**

**300 MHz - 3 GHz**

super high frequency (SHF)

3 GHz - 30 GHz

extremely high frequency (EHF)

30 GHz - 300 GHz

Far infra-red

300 GHz - 3 THz

Heating

3 THz - 30 THz

Light

300 THz

Prodiranje ispod zemljine površine

Talasi vođeni između jonosfere i Zemljine površine

Troposferski i površinski

Jonosferski

Direktni ili prostorni

Apsorpcija od

Apsorpcija od gasa i kise

## Radio Waves



**$10^4 - 10^{-2}$  m/ $10^4 - 10^{10}$  Hz**

ultra-low frequency (ULF)	3 - 30 Hz
extremely low frequency (ELF)	30 - 300 Hz
voice frequencies (VF)	300 Hz - 3 kHz
very low frequency (VLF)	3 - 30 kHz
low frequency (LF)	30 - 300 kHz
medium frequency (MF)	300 kHz - 3 MHz
high frequency (HF)	3 - 30 MHz
<b>very high frequency (VHF)</b>	<b>30 - 300 MHz</b>
<b>ultra high frequency (UHF)</b>	<b>300 MHz - 3 GHz</b>
super high frequency (SHF)	3 - 30 GHz
extremely high frequency (EHF)	30 - 300 GHz
shortwave	MF, HF
<b>television</b>	<b>VHF, UHF</b>
microwave	30 cm - 1 mm/1-300 GHz

**Digitalna TV  
(474 – 858 MHz)**

**Električni talasi**

**Radio talasi**

**Infra crveni**

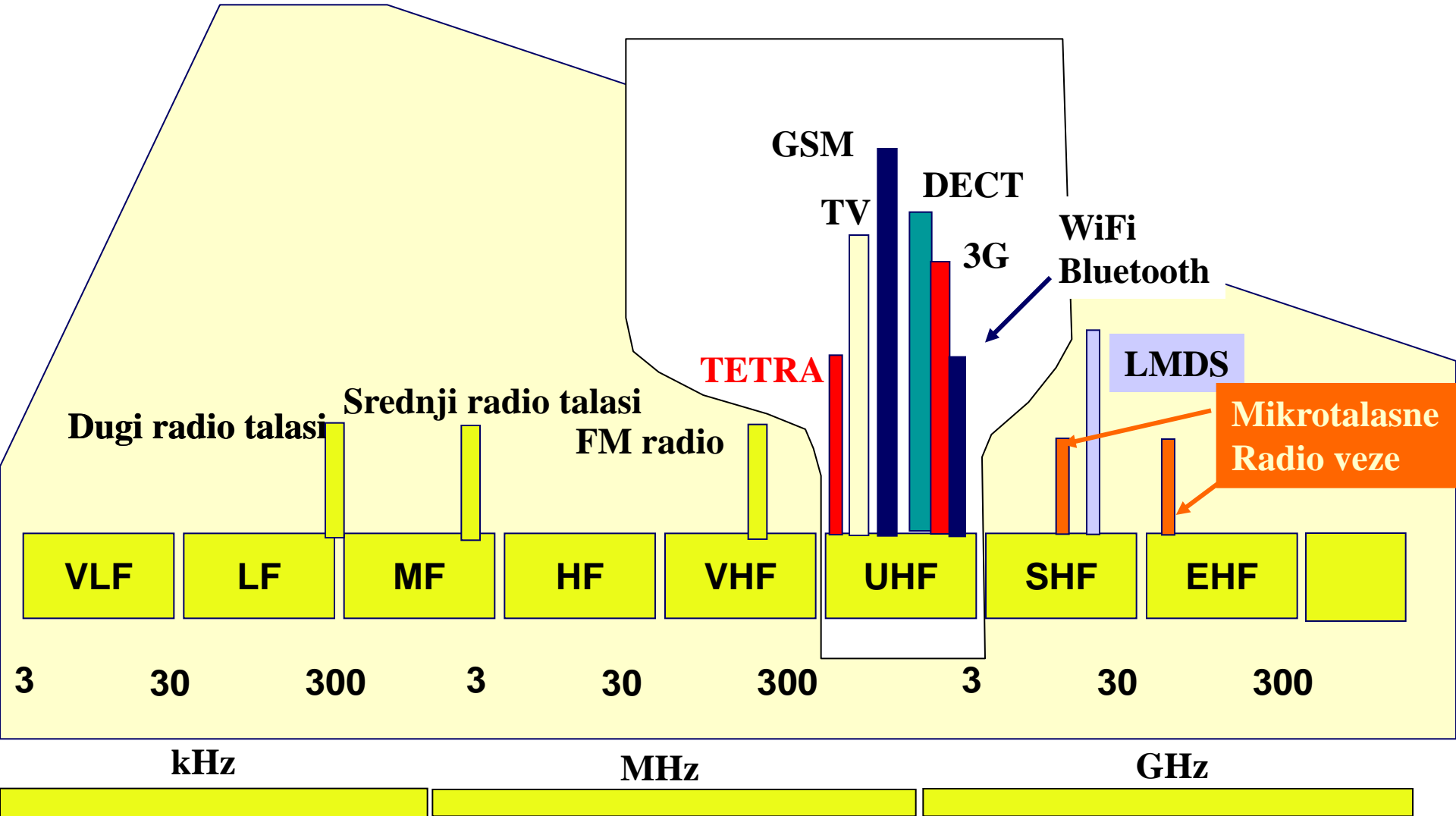
**Vidljiva svetlost**

**Ultra ljubicasti**

**X zraci**

**Gama zraci**

**Kosmički zraci**



**Električni talasi**

**Radio talasi**

**Infra crveni**

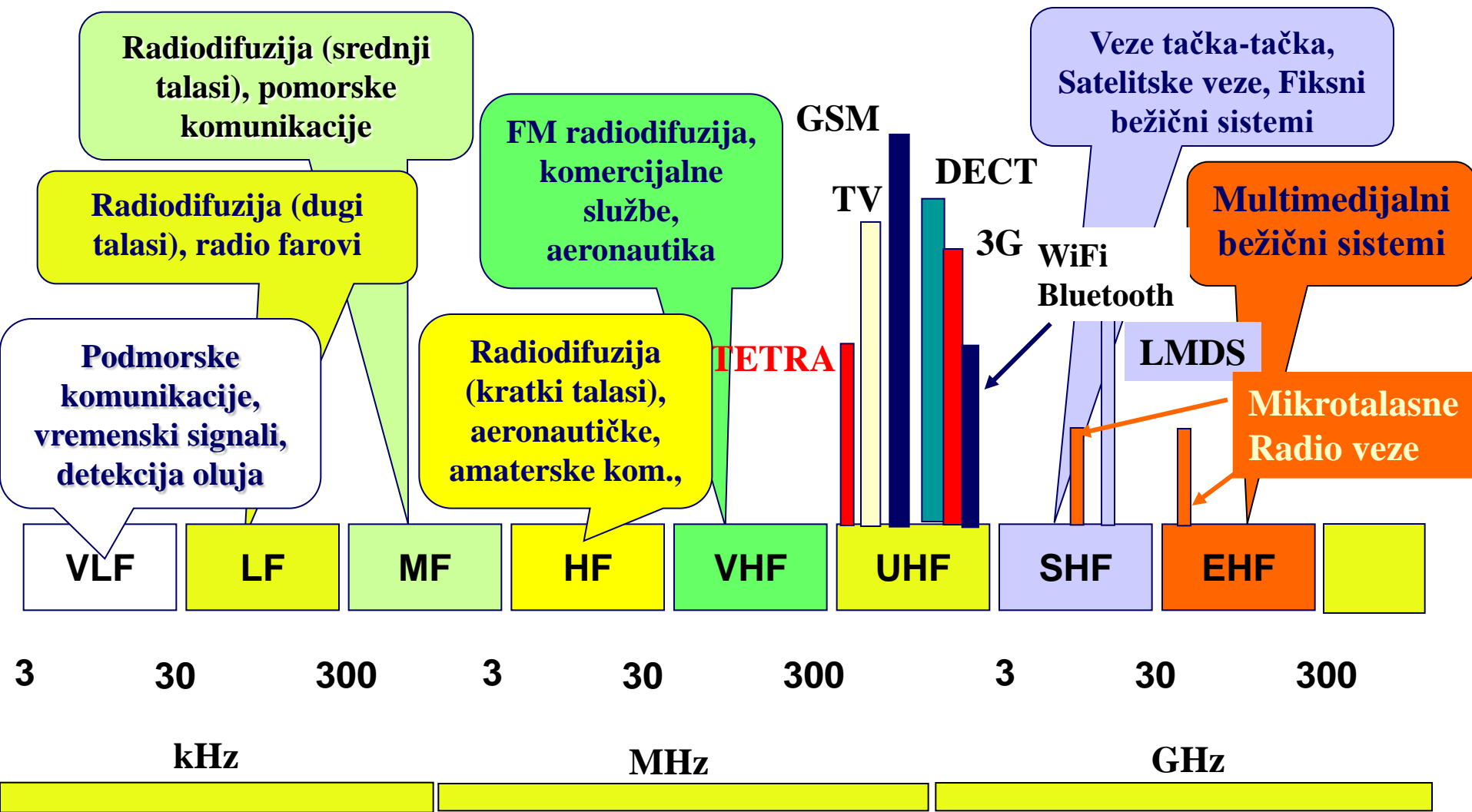
**Vidljiva svetlost**

**Ultra ljubicasti**

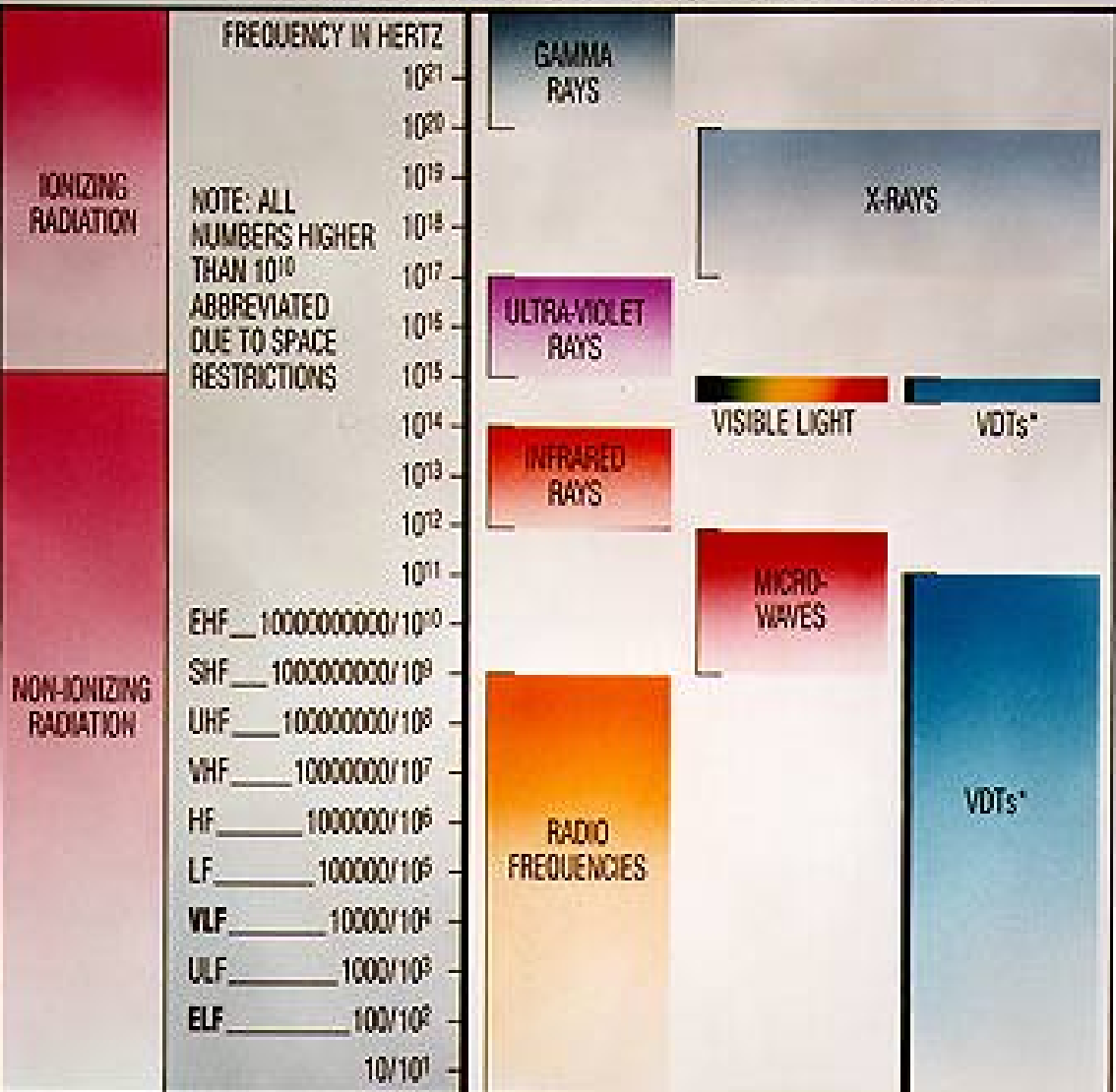
**X zraci**

**Gama zraci**

**Kosmički zraci**



# ELECTROMAGNETIC RADIATION SPECTRUM



\*VDT emissions measured above background noise

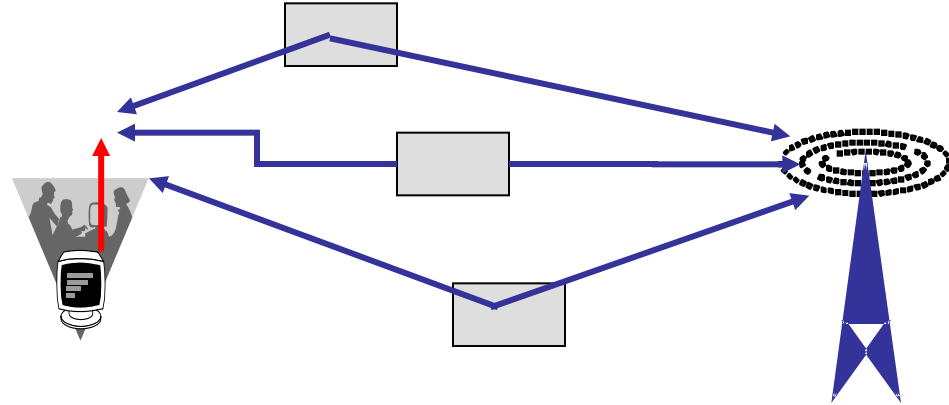
**ABBREVIATIONS:**

- EHF = Extremely High Frequency
- SHF = Super High Frequency
- UHF = Ultra High Frequency
- VHF = Very High Frequency

- HF = High Frequency
- LF = Low Frequency
- VLF = Very Low Frequency
- ULF = Ultra Low Frequency
- ELF = Extremely Low Frequency

# Gubici pri prenosu

- Apsorpcioni gubici
- Uticaj senke
- Višestruka propagacija:
  - i. Fading (Flat fading)
  - ii. Širenje usled Doplerovog efekta
  - iii. Širenje kašnjenja
- Interferencija

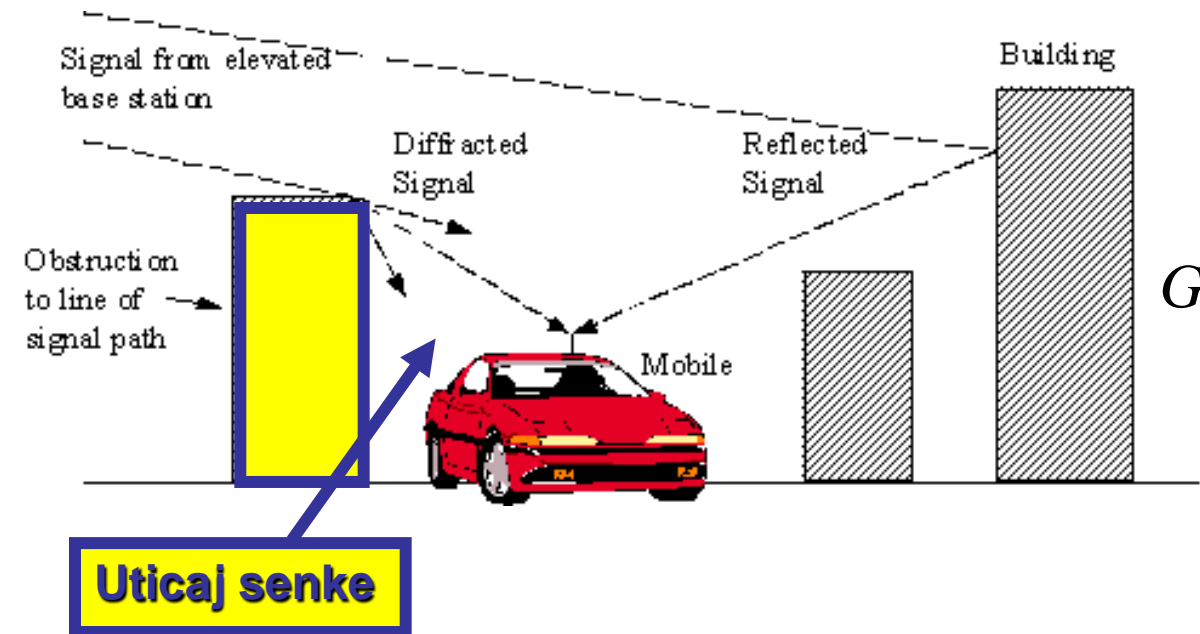


# Apsorpcioni gubici

- Za različita okruženja (planinski predeli, urbane sredine, ..., različiti tipovi drveća, ...) potrebno je koristiti različite modele.
- Jednostavan model gubitaka

$$L = \frac{\bar{P}_r}{P_t} = K \frac{1}{d^\alpha}$$

- $P_r$  lokalna srednja snaga prijemnog signala
- $P_t$  snaga predajnika
- $d$  rastojanje od predajnika
- Eksponent  $\alpha=2$  za slobodan prostor, tipično  $2 \leq \alpha \leq 4$



$$P_r (dB) = \bar{P} (dB) + G_S$$

$$G_S \sim N(0, \sigma_S^2), 4 \leq \sigma_S \leq 10 dB$$

**Senke** nastaju kad postoji prepreka između predajnika i prijemnika, na pr. iza brda, a najgore su u urbanim sredinama (zgrade).

Difrakcija zavisi od frekvencije – izraženija pojava pri niskim frekvencijama.

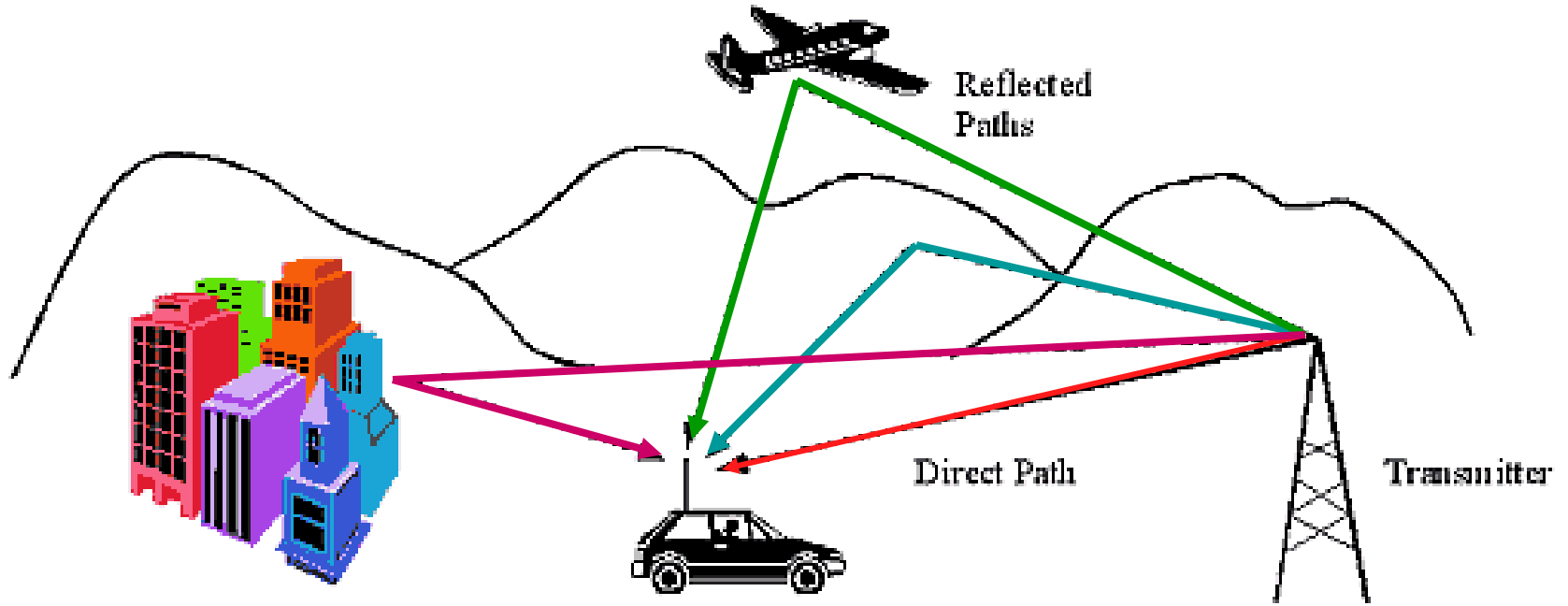
Ovo je uzrok sporog fedinga (log-normalnog fedinga).

Posledica:

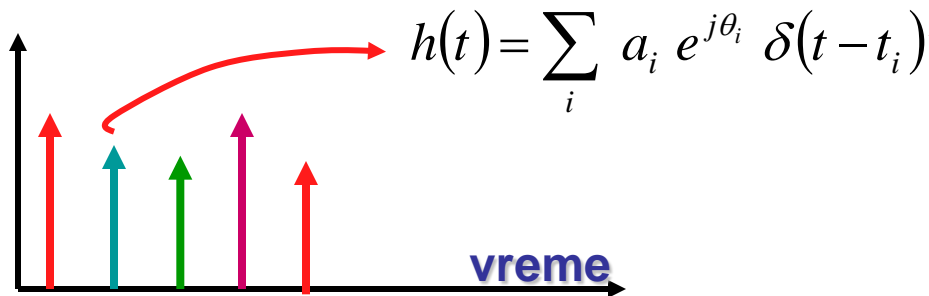
- neuniformno pokrivanje
- povećanje potrebne snage predajnika



# Višestruka propagacija



Prijemna snaga



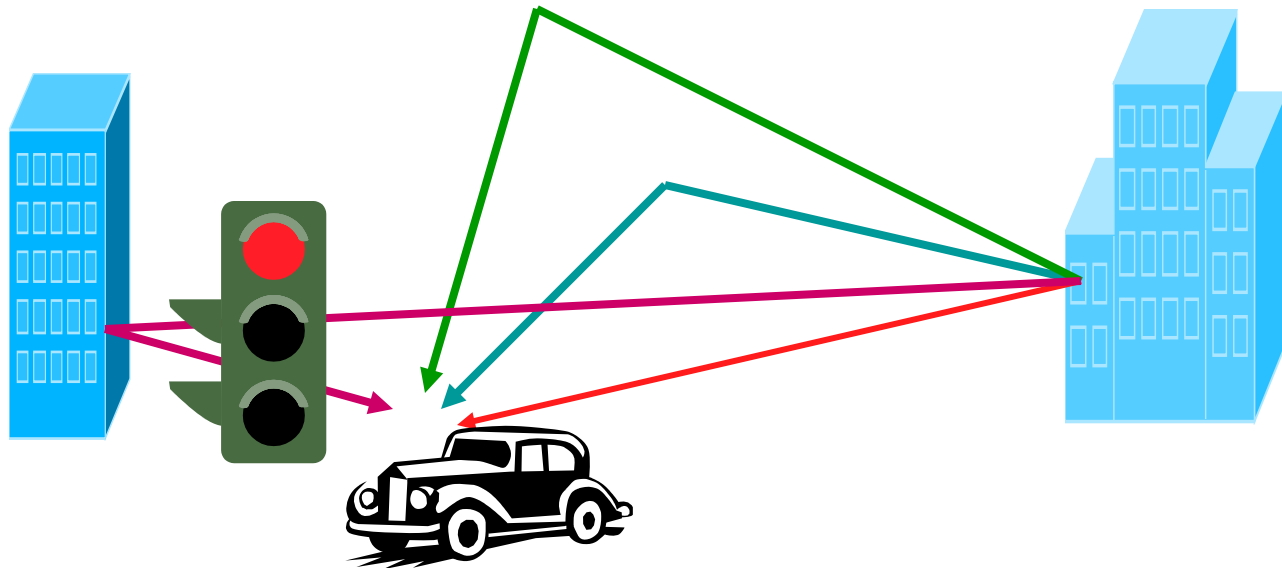
“raspršivanje” kašnjenja

$$h(t) = \sum_i a_i e^{j\theta_i} \delta(t - t_i)$$

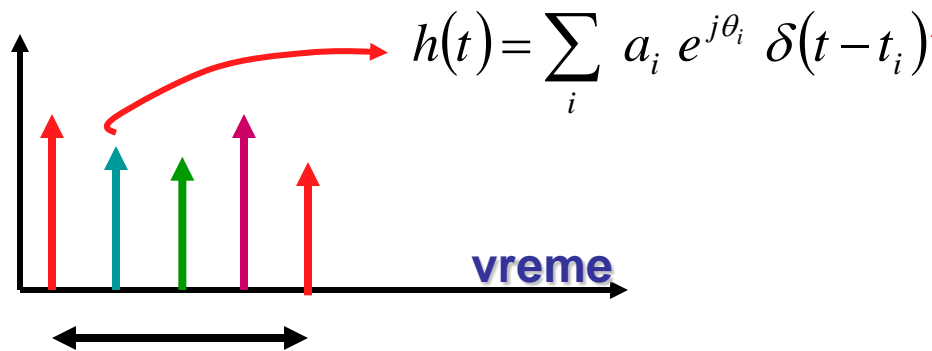
Interferencije:  
•Konstruktivne, i  
•destruktivne



# “red-light effect”



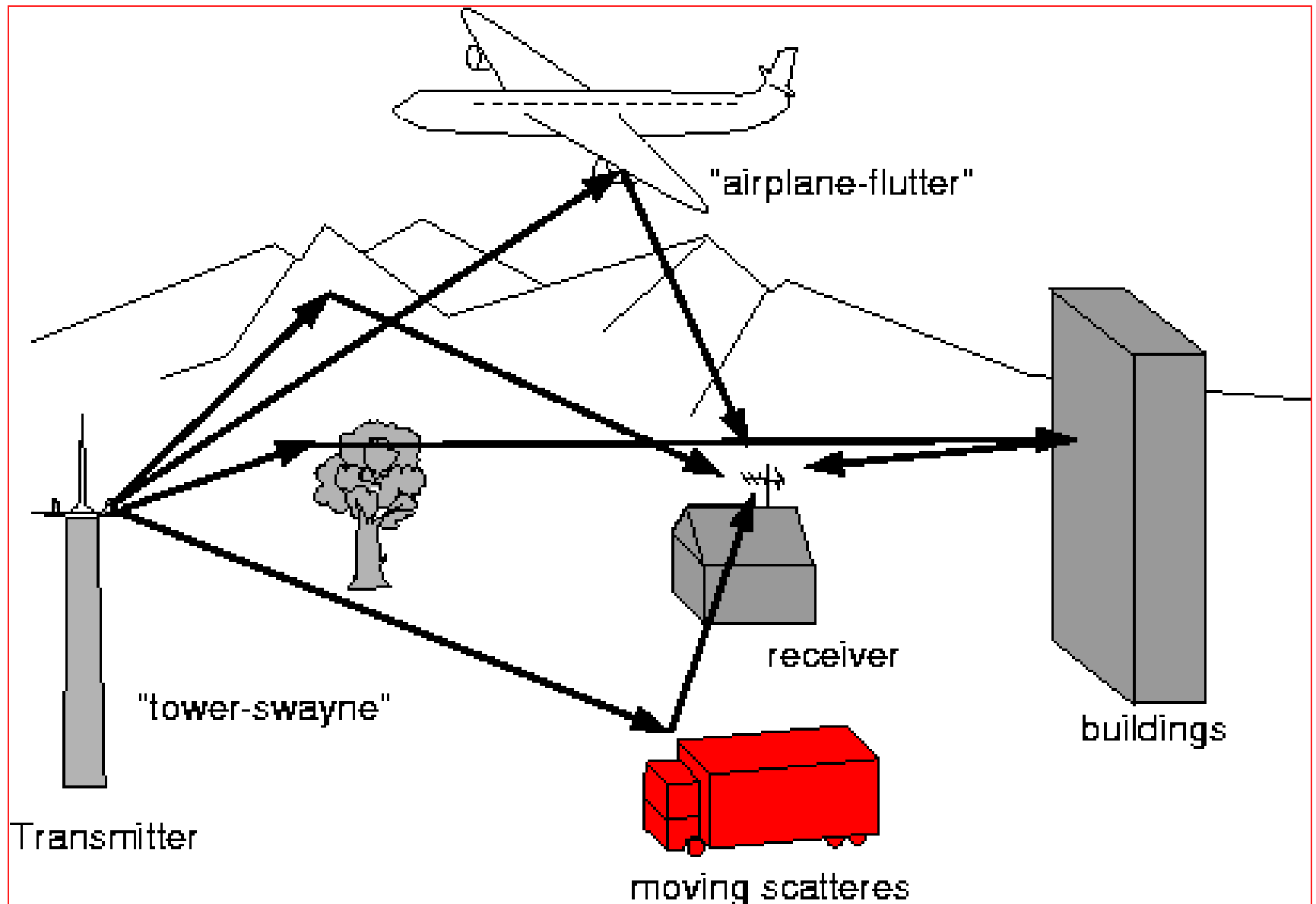
Prijemna snaga



“raspršivanje” kašnjenja

Interferencije:  
•Konstruktivne, i  
•destruktivne



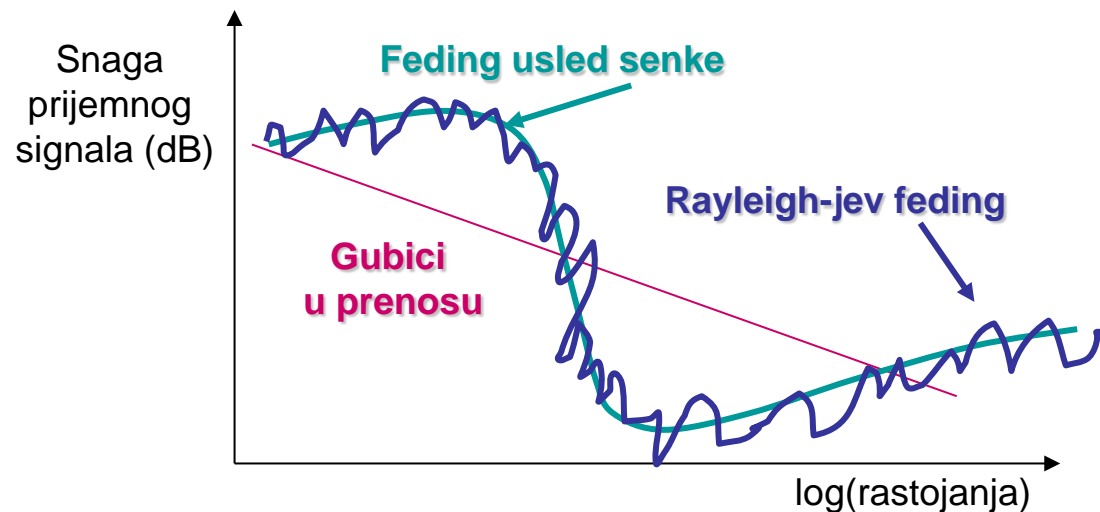


# Feding (flat fading – ravni fading)

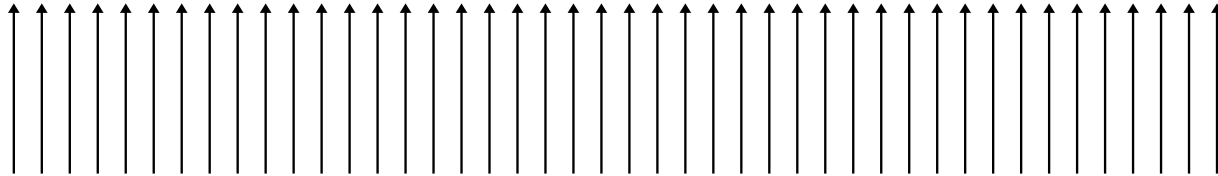
- Ako je raspršivanje kašnjenja malo u poređenju sa periodom simbola
- Anvelopa prijemnog signala,  $r$ , raspodeljena je po Rice-ovoj ili Rayleigh-lijevoj raspodeli

$$P_r (dB) = \bar{P} (dB) + G_s + 20 \log(r)$$

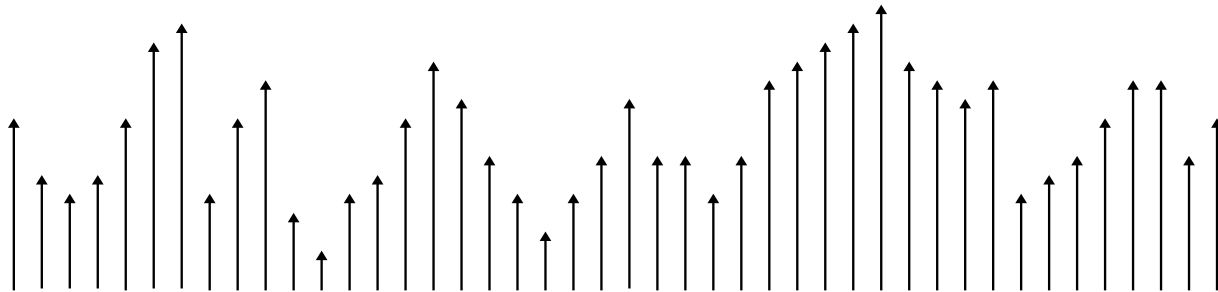
- Posledice:  
povećanje potrebne snage predajnika  
burst-ovi grešaka



# *Uticaj fedinga*



Amplitude nosioca pre fedinga



Posle fedinga

# Širenje usled Doplerovog efekta

$$\Delta f \approx \pm f_0 \frac{v}{c}$$

**Primer:** za  $f_0 = 1\text{GHz}$ , i za  $v = 60\text{km/h}$  (16.7m/s), Doppler-ov pomeraj iznosi:

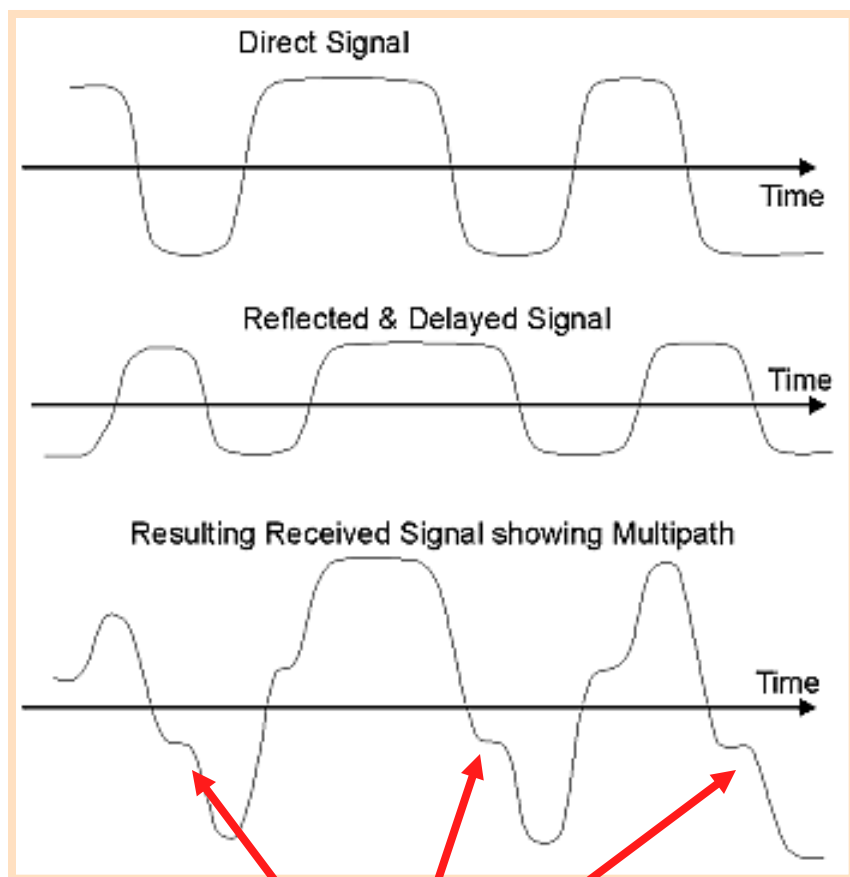
$$\Delta f \approx \pm 10^9 \frac{16.7}{3 \cdot 10^8} = 55.5 \text{ Hz}$$

- Pri približavanju, frekvencija prijemnog signala postaje viša od frekvencije izvora.
- Doplerov efekat utiče na tehnike prenosa osetljive na frekvencijski offset (COFDM), odnosno na one u kojima je velika brzina  $v$  (LEO sateliti).

**Posledica:**

Amplituda i faza su dekorelisane  
posle perioda srazmernog ( $1/f_d$ )

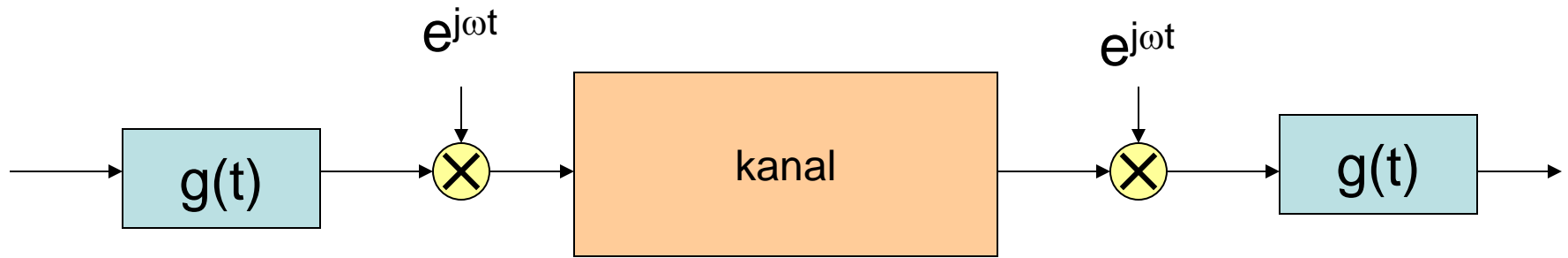
## Širenje kašnjenja – vreme između prve i poslednje značajne komponente signala u višestrukoj propagaciji



**ISI**

Posledice ISI se mogu smanjiti:

- smanjujući protok u svakom kanalu (na pr. podelom opsega pomoću frekvencijskog multipleksiranja, ili OFDM)
- Korišćenjem šema kodovanja koje su manje osetljive na ISI kao što je CDMA.



## Single-carrier transmission

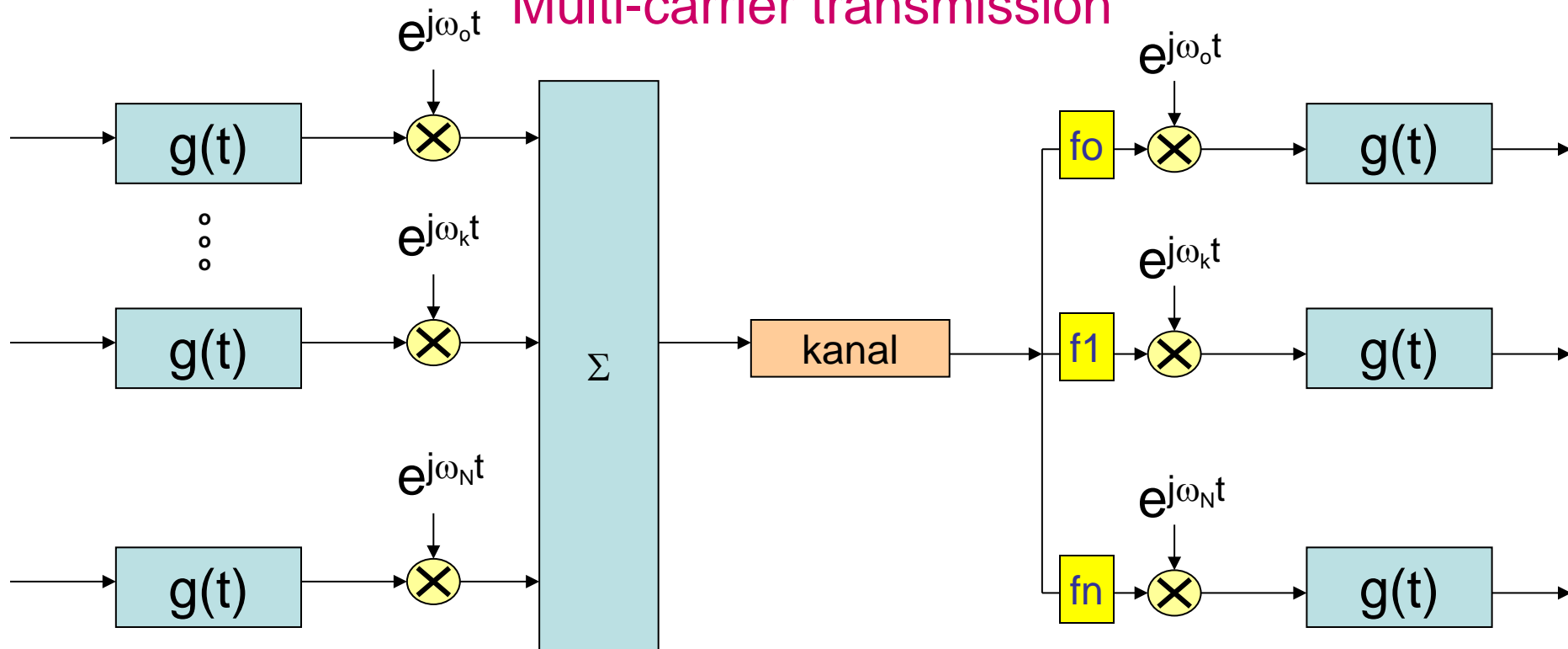
Sa jednim nosiocem, protoci simbola mogu lako dostići 1-30 Msym/s, čemu odgovara perioda simbola reda veličine  $1\mu\text{s}$ . Eho može biti veličine 100-250  $\mu\text{s}$ .

### Ideja je

da se poveća perioda simbola, toliko da smanji ISI, i da se dodaju pauze kao zaštitni vremenski intervali.



## Multi-carrier transmission



- Spektar signala se deli u mnogo uzanih podopsega (potkanala) koji se prenose paralelno
- Idealno-svaki kanal je dovoljno uzan tako da proizvodi samo "flat fading" (pa nema ISI)

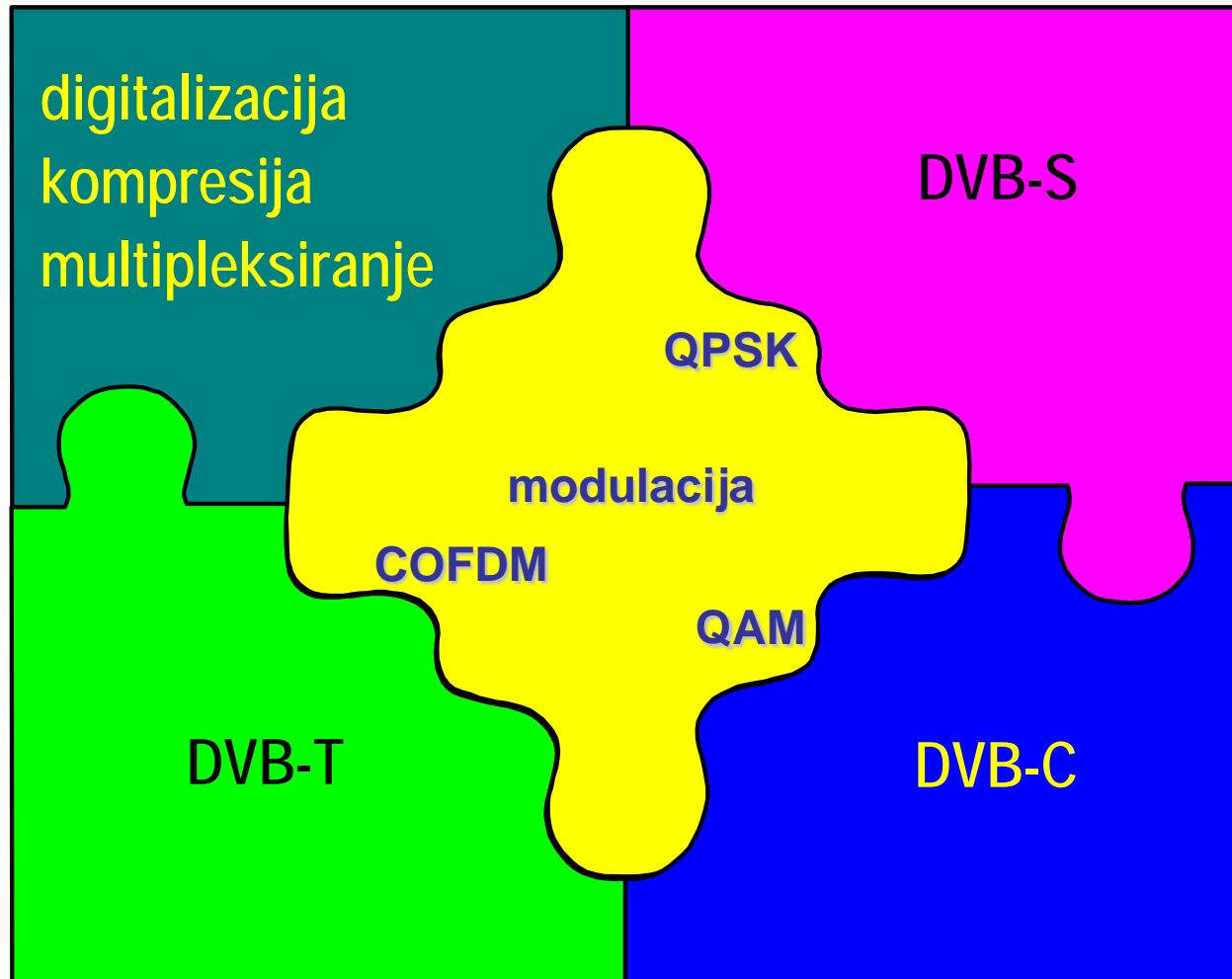
### Nedostaci:

- zahteva se banka filtara na prijemu
- spektralno je neefikasan metod

# Zaključak

Potražiti neku drugu vrstu modulacije  
sa sličnom postavkom

# DVB standard





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Erasmus+ Project No. 561688-EPP-1-2015-1-XK-EPPKA2-CBHE-JP

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**Digital Broadcasting &  
Broadband Technologies**